

### 3A 5V 1MHz Synchronous Buck Converter

### Features

- High Efficiency up to 95%
  Automatic PFM/PWM Mode Operation
- Adjustable Output Voltage from 0.6V to  $V_{PVDD}$
- Integrated 65mW High Side / 55mW Low Side MOSFETs
- Low Dropout Operation: 100% Duty Cycle
- Stable with Low ESR Ceramic Capacitors
- Power-On-Reset Detection on VDD and PVDD
- Integrated Soft-Start and Soft-Stop
- · Over-Temperature Protection
- Over-Voltage Protection
- · Under-Voltage Protection
- High/ Low Side Current Limit
- Power Good Indication
- Enable/Shutdown Function
- TDFN3x3-10 Package
- Lead Free and Green Devices Available
  (RoHS Compliant)

### **Applications**

- Notebook Computer & UMPC
- · LCD Monitor/TV
- Set-Top Box
- · DSL, Switch HUBr
- · Portable Instrument

### **General Description**

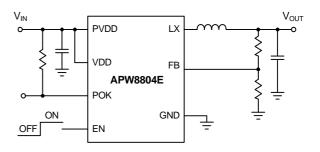
APW8804E is a 3A synchronous buck converter with integrated  $65m\Omega$  high side and  $55m\Omega$  low side power MOSFETs. The APW8804E, design with a current-mode control scheme, can convert wide input voltage of 2.6V to 5.5V to the output voltage adjustable from 0.6V to 5.5V to provide excellent output voltage regulation.

The APW8804E is equipped with an automatic PFM/PWM mode operation. At light load , the IC operates in the PFM mode to reduce the switching losses. At heavy load, the IC works in PWM mode. At PWM mode, the switching frequency is set by the external resistor.

The APW8804E is also equipped with Power-on-reset, soft-start, soft-stop, and whole protections (under-voltage, over-voltage, over-temperature and current-limit) into a single package.

This device, available TDFN3x3-10, provides a very compact system solution external components and PCB area.

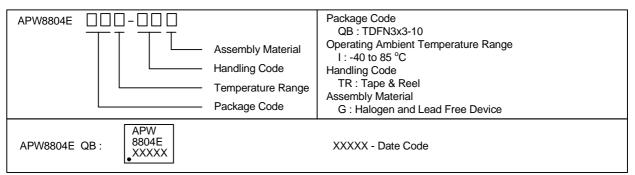
## **Simplified Application Circuit**



ANPEC reserves the right to make changes to improve reliability or manufacturability without notice, and advise customers to obtain the latest version of relevant information to verify before placing orders.

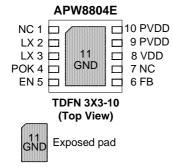


## Ordering and Marking Information



Note: ANPEC lead-free products contain molding compounds/die attach materials and 100% matte tin plate termination finish; which are fully compliant with RoHS. ANPEC lead-free products meet or exceed the lead-free requirements of IPC/JEDEC J-STD-020D for MSL classification at lead-free peak reflow temperature. ANPEC defines "Green" to mean lead-free (RoHS compliant) and halogen free (Br or CI does not exceed 900ppm by weight in homogeneous material and total of Br and CI does not exceed 1500ppm by weight).

## **Pin Configuration**



## Absolute Maximum Ratings (Note 1)

Symbol	Paramet	Rating	Unit	
$V_{\text{PVDD}},V_{\text{VDD}}$	In put Supply Voltage		-0.3 ~ 6.5	V
V	<50ns pulse width		-3 ~V <sub>PVDD</sub> +3	V
V <sub>LX</sub>	LX to GND Voltage >50ns pulse width		-1 ~V <sub>PVDD</sub> +0.3	V
	POK, FB, EN to GND Voltage	-0.3 ~ 6.5	V	
PD	Power Dissipation	2	W	
TJ	Junction Temperature		150	°C
T <sub>STG</sub>	Storage Temperature	-65 ~ 150	°C	
T <sub>SDR</sub>	Maxim um Lead Soldering Temperature, 1	260	°C	

Note1: Stresses beyond those listed under "absolute maximum ratings" may cause permanent damage to the device. These are stress ratings only and functional operation of the device at these or any other conditions beyond those indicated under "recommended operating conditions" is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

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## **Thermal Characteristics**

Symbol	Parameter		Typical Value	Unit
$\theta_{JA}$	Junction-to-Ambient Resistance in Free Air <sup>(Note 2)</sup>	TDFN3x3-10	50	°C/W
θις	Junction-to-Case Resistance in Free Air	TDFN3x3-10	10	°C/W

Note 2:  $\theta_{JA}$  is measured with the component mounted on a high effective thermal conductivity test board in free air. The exposed pad of TDFN3x3-10 is soldered directly on the PCB.

## Recommended Operating Conditions (Note 3)

Symbol	Parameter	Range	Unit
$V_{\text{VDD}}$	Control and Driver Supply Voltage	2.6~ 5.5	V
VPVDD	Input Supply Voltage	2~5.5	V
L	Inductance	1~3.3	μH
I <sub>OUT</sub>	Converter Output Current	0~3	А
T <sub>A</sub>	Am bient Temperature	-40 ~ 85	°C
TJ	Junction Temperature	-40 ~ 125	°C

Note 3: Refer to the typical application circuit.

## **Electrical Characteristics**

Unless otherwise specified, these specifications apply over  $V_{VDD} = V_{PVDD} = 5V$ ,  $T_A = 25^{\circ}C$ .

Cumula a	Parameter Test Conditions	Toot Conditions		APW8804E		11-14
Symbo	Parameter	lest conditions	Min.	Тур.	Max.	Unit
SUPPLY	CURRENT	-			•	•
$I_{VDD}$	VDD Supply Current	V <sub>FB</sub> =0.7V	-	65	-	μΑ
$I_{VDD\_SDH}$	VDD Shutdown Supply Current	EN=GND	-	-	1	μA
POWER-0	DN-RESET (POR)				•	
	VDD POR Voltage Threshold	V <sub>VDD</sub> Rising	2.3	2.4	2.5	V
	VDD POR Hysteresis		-	0.2	-	V
	PVDD POR Voltage Threshold		2.3	2.4	2.5	V
	PVDD POR Hysteresis		-	0.2	-	V
REFEREN	ICE VOLTAGE					
			-	0.6	-	V
Vref	Reference Voltage	All temperature	-1	-	+1	%
	Output Accuracy	I <sub>OUT</sub> =10mA~3A, V <sub>VDD</sub> =2.6~5V	-1.5	-	+1.5	%



## **Electrical Characteristics (Cont.)**

Unless otherwise specified, these specifications apply over V\_{\_{VDD}}=V\_{\_{PVDD}}=5V,\,T\_{\_A}=25^{\circ}C.

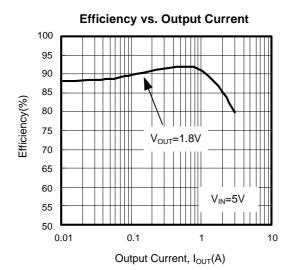
Symbo	Parameter	Test Conditions		AP W880 4E		Unit
Symbo	Parameter	lest conditions	Min.	Тур.	Max.	
OSCILLA	TOR AND DUTY CYCLE				-	
Fosc	Oscillator Frequency		0.85	1	1.15	MHz
	Minimum on Time		-	70	-	ns
POWER	NOSFET					
	High Side P-MOSFET Resistance	V <sub>VDD</sub> =5V, I <sub>LX</sub> =0.5A, T <sub>A</sub> =25°C	-	65	80	mΩ
	Low Side N-MOSFET Resistance	V <sub>VDD</sub> =5V, I <sub>LX</sub> =0.5A, T <sub>A</sub> =25°C	-	55	75	mΩ
	High/Low Side MOSFET Leakage Current		-	-	1	μA
PROTEC	FIONS					
I <sub>LIM</sub>	High Side MOSFET current-limit	Peak Current, V <sub>IN</sub> =2.6∼5.5V T <sub>A</sub> = -40 ~125 °C	4	5	6	A
T <sub>OTP</sub>	Over-tem perature Trip Point (Resoft start after OTP)		-	160	-	°C
	Over-temperature Hysteresis		-	50	-	°C
	Over- Voltage Protection threshold	V <sub>OUT</sub> Rising	120	125	130	%V <sub>RE</sub>
	Under-Voltage Protection threshold		57	66	75	%V <sub>RE</sub>
	Over-Voltage Protection debounce time		-	25	-	μs
	Low Side Switch Current-Limit	From Drain to Source	-	- 1	-	A
SOFT-ST/	ART, ENABLE, AND INPUT CURREN	ITS				
	Soft-Start Time		-	0.8	-	ms
	EN Enable Threshold	$V_{\text{EN}}$ rising voltage to enable device	-	-	1.4	V
	EN Shutdown Threshold	V <sub>EN</sub> falling voltage to shutdown device	0.5	-	-	V
	EN Pull Low Current		-	0.5	1	μΑ
		POK in from Lower (POK Goes High)	82.5	87.5	92.5	%Vou
	POK Threehold	POK Low Hysteresis (POK Goes Low)	-	5	-	%Vol
POK Threshold		POK in from Higher (POK Goes High)	120	125	130	%Vol
		POK High Hysteresis (POK Goes Low)	-	5	-	%V <sub>ou</sub>
	Power Good Pull Low Resistance		-	100	-	Ω
	PowerGoodDebounce	High to low	-	20	-	us



# **Pin Description**

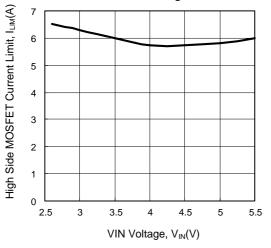
	PIN			
NO.	NAME	FUNCTION		
1	NC	No Connection.		
2,3	LX	Power Switching Output. LX is the Junction of the high-side and low-side Power MOSFETs to supply power to the output LC filter.		
4	РОК	Power Good Output. This pin is open-drain logic output that is pulled to the ground when the output voltage is out of regulation point.		
5	EN	Enable Input. EN is a digital input that turns the regulator on or off. Drive EN high to turn on the regulator, drive it low to turn it off.		
6	FB	Output Feedback Input. The APW8804E senses the feedback voltage via FB and regulates the voltage at 0.6V. Connecting FB with a resistor-divider from the converter's output sets the output voltage.		
7	NC	No connection.		
8	VDD	Signal Input. VDD supplies the control circuitry, gate drivers. Connecting a ceramic bypass capacitor from VDD to GND to eliminate switching noise and voltage ripple on the input to the IC.		
9,10	PVDD	Power Input. PVDD supplies the step-down converter switches. Connecting a ceramic bypass capacitor from PVDD to GND to eliminate switching noise and voltage ripple on the input to the IC.		
11	GND (Exposed Pad)	Ground and Exposed pad. Connect the exposed pad to the system ground plan with large copper area for dissipating heat into the ambient air.		





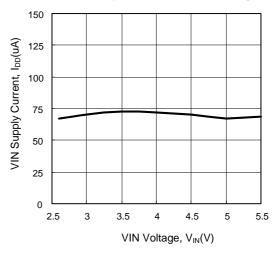
## **Typical Operating Characteristics**





**Reference Voltage vs. VIN Voltage** 0.65 0.64 Reference Voltage, V<sub>REF</sub>(V) 0.63 0.62 0.61 0.6 0.59 0.58 0.57 0.56 0.55 2.5 3 3.5 4 4.5 5 5.5 VIN Voltage, VIN(V)

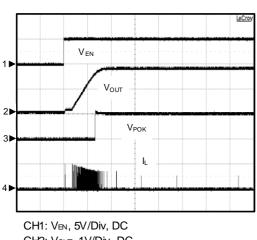
VIN Supply Current vs. VIN Voltage





### **Operating Waveforms**

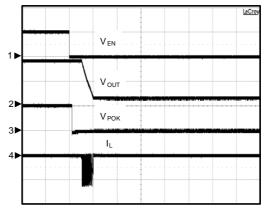
Refer to the typical application circuit. The test condition is  $V_{IN}$ =5V,  $T_A$ = 25°C unless otherwise specified.



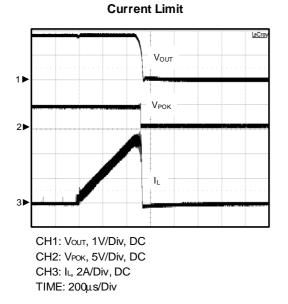
**Enable without Loading** 

СН2: Vovr, 1V/Div, DC CH2: Vovr, 1V/Div, DC CH3: V<sub>POK</sub>, 5V/Div, DC CH4: IL, 1A/Div, DC TIME: 500µs/Div

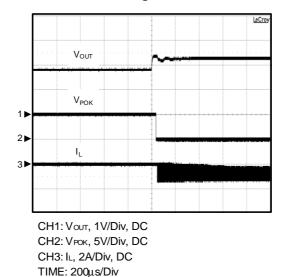
#### Shutdown without Loading



СН1: V<sub>EN</sub>, 5V/Div, DC CH2: Vout, 1V/Div, DC CH3: VPok, 5V/Div, DC CH4: IL, 1A/Div, DC TIME: 200µs/Div



### **Over Voltage Protection**

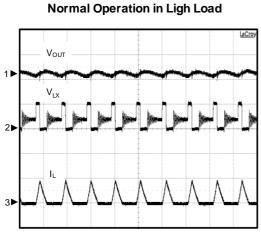


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### **Operating Waveforms**

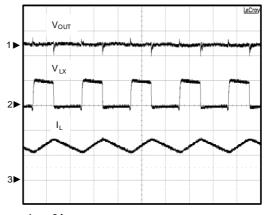
Refer to the typical application circuit. The test condition is  $V_{IN}$ =5V,  $T_A$ = 25°C unless otherwise specified.



lout=200mA

CH1: Vout, 50mV/Div, offset=1.8V CH2: VLx, 5V/Div, DC CH3: IL, 1A/Div, DC TIME: 2µs/Div

### Normal Operation in Heavy Load

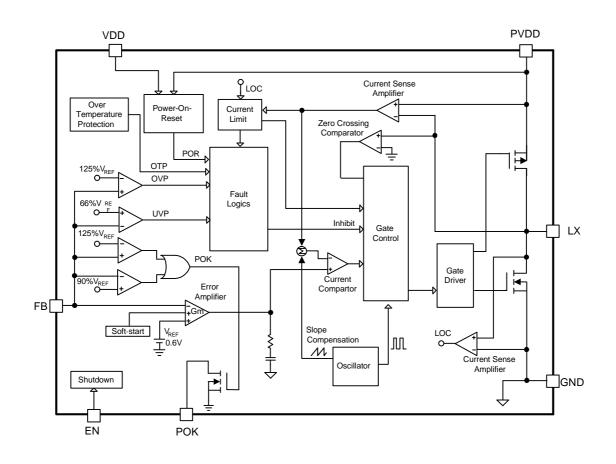


Iout=3A

CH1: Vouт, 50mV/Div, offset=1.8V CH2: VLx, 5V/Div, DC CH3: IL, 2A/Div, DC TIME: 0.5µs/Div

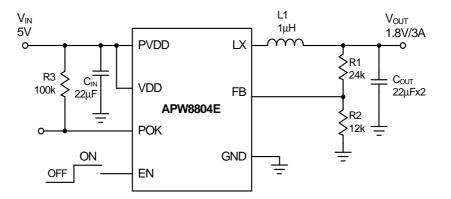


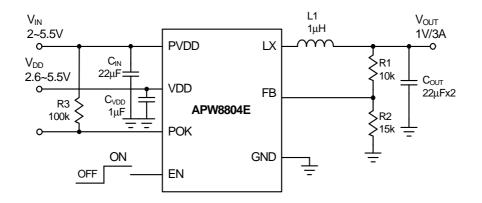
## **Block Diagram**





# **Typical Application Circuit**





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### **Function Description**

#### VDD and PVDD Power-On-Reset (POR)

The APW8804E keeps monitoring the voltage on VDD and PVDD pins to prevent wrong logic operations which may occur when VDD or PVDD voltage is not high enough for internal control circuitry to operate. The VDD POR rising threshold is 2.4V (typical) with 0.2V hysteresis and PVDD POR rising threshold is 1.7V with 0.2V hysteresis. During start-up, the VDD and PVDD voltage must exceed the POR threshold. Then, the IC starts a start-up process and ramps up the output voltage to the voltage target.

#### **Output Under-Voltage Protection (UVP)**

In the operational process, if a short-circuit occurs, the output voltage will drop quickly. Before the current-limit circuit responds, the output voltage will fall out of the required regulation range. The under-voltage continually monitors the FB voltage after soft-start is completed. If a load step is strong enough to pull the output voltage lower than the under-voltage threshold, the IC shuts down converter's output.

The under-voltage threshold is 66% of the nominal output voltage. APW8804E will be latched after under-voltage protection.

#### **Over-Voltage Protection (OVP)**

The over-voltage function monitors the output voltage by FB pin. When the FB voltage increases over 125% of the reference voltage due to the high-side MOSFET failure or for other reasons, the over-voltage protection comparator will trigger soft-stop function and shutdown the converter output.

#### **Over-Temperature Protection (OTP)**

The over-temperature circuit limits the junction temperature of the APW8804E. When the junction temperature exceeds  $T_J$ =+160°C, a thermal sensor turns off the both power MOSFETs, allowing the devices to cool. The thermal sensor allows the converters to start a start-up process and to regulate the output voltage again after the junction temperature cools by 50°C. The OTP is designed

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with a 50°C hysteresis to lower the average  $T_J$  during continuous thermal overload conditions, increasing life-time of the APW8804E.

#### **Current-Limit Protection**

The APW8804E monitors the output current, flows through the high-side and low-side power MOSFETs, and limits the current peak at current-limit level to prevent the IC from damaging during overload, short-circuit and overvoltage conditions. Typical high side power MOSFET current limit is 5A, and low side MOSFET current limit is -1A.

#### Soft-Start

The APW8804E has a built-in soft-start to control the rise rate of the output voltage and limit the input current surge during start-up. During soft-start, an internal voltage ramp connected to one of the positive inputs of the error amplifier, rises up to replace the reference voltage (0.6V) until the voltage ramp reaches the reference voltage. During soft-start without output over-voltage, the APW8804E converter's sinking capability is disabled until the output voltage reaches the voltage target.

#### Soft-Stop

At the moment of shutdown controlled by EN signal or over-voltage event, the APW8804E initiates a soft-stop process to discharge the output voltage in the output capacitors. Certainly, the load current also discharges the output voltage. During soft-stop, the low side MOSFET turns on each cycle to discharge the output voltage. Therefore, the output voltage falls down slowly at the light load. After the soft-stop interval elapses, the soft-stop process ends and the IC turns off.

#### **Enable and Shutdown**

Driving EN to ground places the APW8804E in shutdown. In shutdown mode, the internal power MOSFETs turn off, all internal circuitry shuts down and the quiescent supply current reduces to less than  $1\mu$ A.



## Function Description (Cont.)

### **Powr Good Indicator**

POK is actively held low in shutdown and soft-start status. In the soft-start process, the POK is an open-drain. When the soft-start is finished, the POK is released. In normal operation, the POK window is from 85% to 125% of the converter reference voltage. When the output voltage stays within this window, POK signal will become high. When the output voltage outruns 80% or 125% of the target voltage, POK signal will be pulled low immediately. In order to prevent false POK drop, capacitors need to parallel at the output to confine the voltage deviation with severe load step transient.



### **Application Information**

#### Input Capacitor Selection

Because buck converters have a pulsating input current, a low ESR input capacitor is required. This results in the best input voltage filtering, minimizing the interference with other circuits caused by high input voltage spikes. Also, the input capacitor must be sufficiently large to stabilize the input voltage during heavy load transients. For good input voltage filtering, usually a  $22\mu$ F input capacitor is sufficient. It can be increased without any limit for better input-voltage filtering. Ceramic capacitors show better performance because of the low ESR value, and they are less sensitive against voltage transients and spikes compared to tantalum capacitors. Place the input capacitor as close as possible to the input and GND pin of the device for better performance.

#### **Inductor Selection**

For high efficiencies, the inductor should have a low DC resistance to minimize conduction losses. Especially at high-switching frequencies, the core material has a higher impact on efficiency. When using small chip inductors, the efficiency is reduced mainly due to higher inductor core losses. This needs to be considered when selecting the appropriate inductor. The inductor value determines the inductor ripple current. The larger the inductor value, the smaller the inductor ripple current and the lower the conduction losses of the converter. Conversely, larger inductor values cause a slower load transient response. A reasonable starting point for setting ripple current,  $\Delta I_{L}$  is 40% of maximum output current. The recommended inductor value can be calculated as below:

$$L \geq \frac{V_{OUT} \left(1 - \frac{V_{OUT}}{V_{IN}}\right)}{F_{SW} \cdot \Delta I_{L}}$$

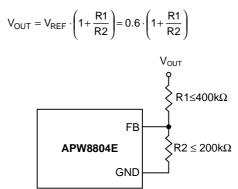
$$\mathbf{I}_{\text{L(MAX)}} = \mathbf{I}_{\text{OUT(MAX)}} + 1/2 \times \Delta \mathbf{I}_{\text{L}}$$

To avoid the saturation of the inductor, the inductor should be rated at least for the maximum output current of the converter plus the inductor ripple current.

#### **Output Voltage Setting**

In the adjustable version, the output voltage is set by a resistive divider. The external resistive divider is connected to the output, allowing remote voltage sensing as

Copyright © ANPEC Electronics Corp. Rev. A.3 - Jul., 2017 shown in "Typical Application Circuits". A suggestion of maximum value of R2 is  $200k\Omega$  to keep the minimum current that provides enough noise rejection ability through the resistor divider. The output voltage can be calculated as below:

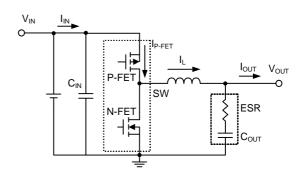


#### **Output Capacitor Selection**

The current-mode control scheme of the APW8804E allows the use of tiny ceramic capacitors. The higher capacitor value provides the good load transients response. Ceramic capacitors with low ESR values have the lowest output voltage ripple and are recommended. If required, tantalum capacitors may be used as well. The output ripple is the sum of the voltages across the ESR and the ideal output capacitor.

$$\Delta V_{OUT} \cong \frac{V_{OUT} \cdot \left(1 - \frac{V_{OUT}}{V_{IN}}\right)}{F_{SW} \cdot L} \cdot \left(ESR + \frac{1}{8 \cdot F_{SW} \cdot C_{OUT}}\right)$$

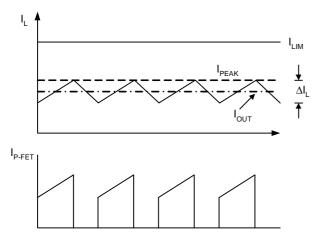
When choosing the input and output ceramic capacitors, choose the X5R or X7R dielectric formulations. These dielectrics have the best temperature and voltage characteristics of all the ceramics for a given value and size.





## Application Information (Cont.)

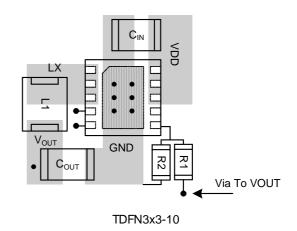
#### **Output Capacitor Selection (Cont.)**

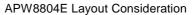


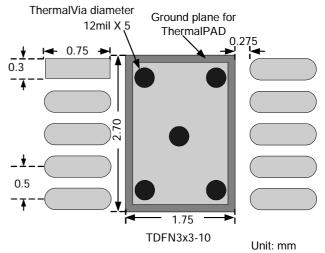
#### Layout Consideration

For all switching power supplies, the layout is an important step in the design; especially at high peak currents and switching frequencies. If the layout is not carefully done, the regulator might show noise problems and duty cycle jitter.

- The input capacitor should be placed close to the PVDD and GND. Connecting the capacitor and PVDD/GND with short and wide trace without any via holes for good input voltage filtering. The distance between VIN/GND to capacitor less than 2mm respectively is recommended.
- To minimize copper trace connections that can inject noise into the system, the inductor should be placed as close as possible to the LX pin to minimize the noise coupling into other circuits.
- 3. The output capacitor should be place closed to LX and GND.
- 4. Since the feedback pin and network is a high impedance circuit the feedback network should be routed away from the inductor. The feedback pin and feedback network should be shielded with a ground plane or trace to minimize noise coupling into this circuit.
- 5. A star ground connection or ground plane minimizes ground shifts and noise is recommended.







APW8804E Recommended Footprint

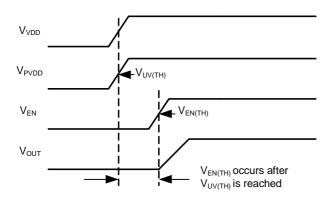


## **Application Information (Cont.)**

#### **Input Power Sequence**

The APW8804E has three power-on control pins, PVDD, VDD and EN input pins. It is recommended that the three inputs should have a standard sequence. The VDD and PVDD input powers should be present to their respective pins regardless of input sequence. After VDD and PVDD powers are ready, the enable signal can be present to EN pin and the IC will be enabled.

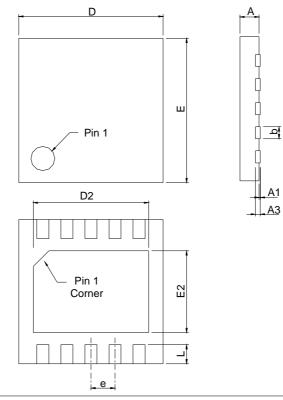
Figure 1 shows the VDD and PVDD comes up before the VEN. Recommended power on sequence is shown in Figure1.





## Package Information

TDFN3x3-10

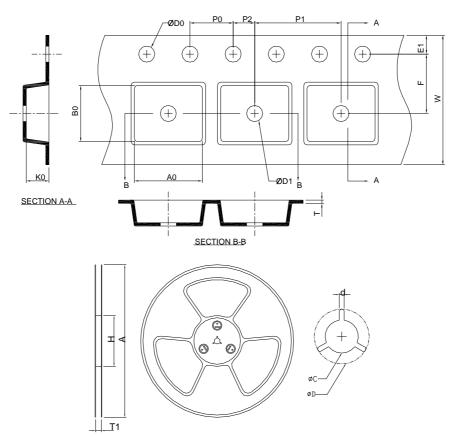


Ş		TDFN	3x3-10	
SY-MBOL	MILLIM	ETERS	INC	HES
Ē	MIN.	MAX.	MIN.	MAX.
A	0.70	0.80	0.028	0.031
A1	0.00	0.05	0.000	0.002
A3	0.20 REF		0.00	8 REF
b	0.18	0.30	0.007	0.012
D	2.90	3.10	0.114	0.122
D2	2.20	2.70	0.087	0.106
E	2.90	3.10	0.114	0.122
E2	1.40	1.75	0.055	0.069
е	0.50	BSC	0.02	0 BSC
L	0.30	0.50	0.012	0.020
К	0.20		0.008	

Note : 1. Followed from JEDEC MO-229 VEED-5.



# **Carrier Tape & Reel Dimensions**



Application	A	Н	T1	С	d	D	W	E1	F
	330.0±2.00	50 MIN.	12.4+2.00 -0.00	13.0+0.50 -0.20	1.5 MIN.	20.2 MIN.	12.0±0.30	1.75±0.10	5.5±0.05
TDFN3x3-10	P0	P1	P2	D0	D1	Т	A0	B0	K0
	4.0±0.10	8.0±0.10	2.0±0.05	1.5+0.10 -0.00	1.5 MIN.	0.6+0.00 -0.40	3.30±0.20	3.30±0.20	1.30±0.20

(mm)

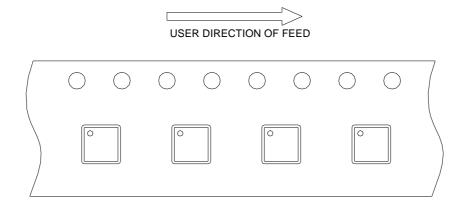
## **Devices Per Unit**

Package Type	Unit	Quantity
TDFN3x3-10	Tape & Reel	3000

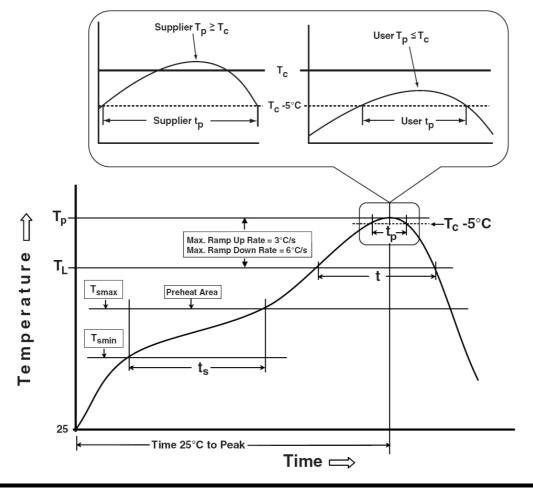


## **Taping Direction Information**

TDFN3x3-10



## **Classification Profile**



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### **Classification Reflow Profiles**

Profile Feature	Sn-Pb Eutectic Assembly	Pb-Free Assembly			
$\begin{array}{c} \textbf{Preheat \& Soak} \\ \textbf{Temperature min (T_{smin})} \\ \textbf{Temperature max (T_{smax})} \\ \textbf{Time (T_{smin} to T_{smax}) (t_s)} \end{array}$	100 °C 150 °C 60-120 seconds	150 °C 200 °C 60-120 seconds			
Average ramp-up rate (T <sub>smax</sub> to T <sub>P</sub> )	3 °C/second max.	3°C/second max.			
Liquidous temperature $(T_L)$ Time at liquidous $(t_L)$	183 °C 60-150 seconds	217 °C 60-150 seconds			
Peak package body Temperature (T <sub>p</sub> )*	See Classification Temp in table 1	See Classification Temp in table 2			
Time $(t_P)^{**}$ within 5°C of the specified classification temperature $(T_c)$	20** seconds	30** seconds			
Average ramp-down rate ( $T_p$ to $T_{smax}$ )	6 °C/second max.	6 °C/second max.			
Time 25°C to peak temperature	6 minutes max.	8 minutes max.			
* Tolerance for peak profile Temperature ( $T_p$ ) is defined as a supplier minimum and a user maximum. ** Tolerance for time at peak profile temperature ( $t_p$ ) is defined as a supplier minimum and a user maximum.					

#### Table 1. SnPb Eutectic Process – Classification Temperatures (Tc)

Package Thickness	Volume mm <sup>3</sup> <350	Volume mm <sup>3</sup> <sup>3</sup> 350
<2.5 mm	235 °C	220 °C
≥2.5 mm	220 °C	220 °C

Table 2. Pb-free Process – Classification Temperatures (Tc)

Package Thickness	Volume mm <sup>3</sup> <350	Volume mm <sup>3</sup> 350-2000	Volume mm <sup>3</sup> >2000
<1.6 mm	260 °C	260 °C	260 °C
1.6 mm – 2.5 mm	260 °C	250 °C	245 °C
≥2.5 mm	250 °C	245 °C	245 °C

## **Reliability Test Program**

Test item	Method	Description
SOLDERABILITY	JESD-22, B102	5 Sec, 245°C
HOLT	JESD-22, A108	1000 Hrs, Bias @ 125°C
PCT	JESD-22, A102	168 Hrs, 100%RH, 2atm, 121°C
ТСТ	JESD-22, A104	500 Cycles, -65°C~150°C
НВМ	MIL-STD-883-3015.7	VHBM≧2KV
MM	JESD-22, A115	VMM≧200V
Latch-Up	JESD 78	10ms, $1_{tr} \ge 100 \text{mA}$



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