Datasheet

## FS98002

8-bit MCU with 2k program EPROM, 128-byte RAM, 14-bit ADC, 12-bit GIO, $4 \times 12$ LCD driver

## Fortune Semiconductor Corporation

富晶電子股份有限公司
23F．，No．29－5，Sec．2，Zhongzheng E．Rd．，
Danshui Dist．，New Taipei City 251，Taiwan
Tel．：886－2－28094742
Fax ：886－2－28094874
www．ic－fortune．com

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## 1. General Description

The FS98O02 is a high performance, low cost 8 -bit MCU with 2 k program EPROM, 128-byte data RAM, one 14-bit ADC, 12-bit GIO, and $4 \times 12$ LCD driver. With a few external passive components such as resistors and capacitors, the FS98O02 can be easily implemented to form a simple portable tire gauge, voltage panel meter, or manual range DMM, etc. Especially, FS98O02's EPROM could be written by program instruction, so users can write table or calibration data in the unused EPROM address more than one time.

## 2. Features

- 8-bit RISC CPU core with 39 single word instructions.
- Embedded $2 \mathrm{k} \times 16$ program ROM ( 07 ffH isn't usable by user), 128-byte data RAM.
- Instruction Programmer Function.
- Operating voltage is from 2.4 V to 3.6 V .
- Operating current is about 1.5 mA ; sleep current is about $2 \mu \mathrm{~A}$.
- Embedded internal $1 \mathrm{MHz} / 4 \mathrm{MHz}$ oscillator.
- 4-level hardware stacks.
- 3 Interrupt sources (external: input port $1<0>$, internal: timer, ADC)
- One 14-bit noise free ADC.
- Embedded Voltage Regulator ( $2.5 \mathrm{~V} / 3.6 \mathrm{~V}$ regulated output for FS98002A and FS98002C).
- Embedded Voltage Regulator ( $2.5 \mathrm{~V} / \mathrm{VDD}_{(\text {max. } 3.6 \mathrm{~V})}$ regulated output for FS98002B $)$
- Embedded Voltage Doubler (3V/2xVDDP pumped output).
- 4-bit Input Port and 8-bit bi-directional I/O port including 1-bit for buzzer output.
- Embedded Low Voltage Reset (LVR) and Low Battery Detector (LBD) .
- $4 \times 12$ LCD driver (3V peak-to-peak).
- Watchdog timer.


## 3. Applications

- Simple portable tire gauge.
- Voltage panel meter.
- Manual range DMM.
- Scale.


## 4. Ordering Information

| Product Number | Description | Package Type |
| :--- | :--- | :--- |
| FS98O02A-D | MCU with OTP ROM ; The customer has to program the compiled <br> hex code into OTP ROM. | 49-pin Dice form |
| FS98O02A-nnnV-D | MCU with program type ; FSC programs the customer's compiled <br> hex code into OTP ROM at factory before shipping. | 49-pin Dice form |
| FS98O02A-PCE | MCU with OTP ROM ; The customer has to program the compiled <br> hex code into OTP ROM. | 64-pin LQFP |
| FS98O02A-nnnV-PCE | MCU with program type ; FSC programs the customer's compiled <br> hex code into OTP ROM at factory before shipping. | 64-pin LQFP |
| FS98O02B-D | MCU with OTP ROM ; The customer has to program the compiled <br> hex code into OTP ROM. | 49-pin Dice form |
| FS98O02B-nnnV-D | MCU with program type ; FSC programs the customer's compiled <br> hex code into OTP ROM at factory before shipping. | 49-pin Dice form |
| FS98O02B-PCE | MCU with OTP ROM ; The customer has to program the compiled <br> hex code into OTP ROM. | 64-pin LQFP |
| FS98O02B-nnnV-PCE | MCU with program type ; FSC programs the customer's compiled <br> hex code into OTP ROM at factory before shipping. | 64-pin LQFP |
| FS98O02C-D | MCU with OTP ROM ; The customer has to program the compiled <br> hex code into OTP ROM. | 49-pin Dice form |
| FS98O02C-nnnV-D | MCU with program type ; FSC programs the customer's compiled <br> hex code into OTP ROM at factory before shipping. | 49-pin Dice form |
| FS98O02C-PCE | MCU with OTP ROM ; The customer has to program the compiled <br> hex code into OTP ROM. | 64-pin LQFP |
| FS98O02C-nnnV-PCE | MCU with program type ; FSC programs the customer's compiled <br> hex code into OTP ROM at factory before shipping. | 64-pin LQFP |

Note1 : Code number ( nnnV ) is assigned for customer.
Note2 : Code number ( $n n n=001 \sim 999$ ) ; Version ( $\mathrm{V}=\mathrm{A} \sim \mathrm{Z}$ )
Note3 : PCE means package of Pb-free and LQFP 64 pin.

## 5. Pin Configuration



Figure 5-1: LQFP64

## 6. Pin Description

| Name | In/Out | Pin No | Description |
| :---: | :---: | :---: | :---: |
| P1<0>/INT | I | 3 | Input port 1.0 or interrupt input |
| P1<1>/PROEN | I | 4 | Input port1.1, SPI or Instruction Programmer select |
| P1<2>~P1<3> | I | 5~6 | Input port 1.2~1.3 |
| $\mathrm{P} 1<4>\sim \mathrm{P} 1<5>$ | I/O | 7~8 | I/O port 1.4~1.5 |
| P1<6>/ Analog Input | I/O | 9 | I/O port 1.6 or Analog channel input |
| PT1<7>/BZ | I/O | 10 | I/O port 1.7 or buzzer output |
| PT2<0>~PT2<3> | I/O | 11~14 | I/O port 2.0~2.3 |
| SEG12~SEG1 | O | 15~26 | LCD segment driver output |
| COM4~COM3 | 0 | 28~29 | LCD common driver output |
| COM2~COM1 | 0 | 35~36 | LCD common driver output |
| RLCD | 1 | 37 | LCD Voltage Input (usually connect 10~100k 施 VDDA or 3V VGG ) |
| VS | O | 38 | Voltage source from VDDA |
| VDDA | 0 | 39 | Voltage regulator power output (2.5V / 3.6V for FS98O02A, FS98002C or $2.5 \mathrm{~V} / \mathrm{VDD}_{(\text {max. } 3.6 \mathrm{~V})}$ for FS98O02B ) |
| VGG | O | 41 | Voltage doubler output ( 3 V or $2 \times \mathrm{VDDP}$ ) |
| TST | I | 42 | Test Mode control pin (low active) |
| CA | I/O | 43 | Voltage doubler capacitor positive connection |


| Name | In/Out | Pin No | Description |
| :--- | :--- | :--- | :--- |
| CB | I/O | 44 | Voltage doubler capacitor negative connection |
| VDDP | I | 45 | Analog power supply |
| VDD | I | 47 | Positive power supply |
| VSS | I | 48 | Negative power supply ( ground ) |
| VB | I | 51 | Analog circuit bias current input |
| AGND | I/O | 52 | Analog ground |
| OP1O | O | 53 | OPAMP output |
| AIN0(AD0)~ AIN1(AD1) | I | $55 \sim 56$ | Analog signal input channel |
| AIN2(AD2) | I | 57 | Analog signal input channel (usually for ADC VIL input) |
| AIN3(AD3) | I | 58 | Analog signal input channel (usually for ADC VRH input) |
| AIN4(AD4) | I | 59 | Analog signal input channel (usually for ADC VRL input) |
| FTB, FTC | I/O | 61,62 | ADC pre-filter capacitor connection |
| VPP/RST | I | 64 | Program Input Voltage or Reset |

7. Functional Block Diagram


Figure 7-1: Functional Block Diagram
8. Typical Application Circuit


Figure 8-1: Scale, Instruction Programmer Calibration Data to EPROM

Note. In the instruction program mode, VPP must be connected to $100 \Omega$ Resistor. Please keep VGG between 5.4 and 6.2 V when executing instruction programming. Please turn off VDDA or VS before executing instruction programming if the loading of VDDA or VS is over 8 mA . And turn off the LCD bias circuit in the instruction program mode if the application circuit is VGG pin connect to RLCD pin.


Figure 8-2: Scale, Normal mode for FS98002A and FS98002B
 operation can be selected part B. The resistor of the 100k Ohm is used to limit the current

Figure 8-3: Scale, Normal mode for FS98002C


Figure 8-4: Tire Gauge

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## 9. Electrical Characteristics

### 9.1 Absolute Maximum Ratings

| Parameter | Rating | Unit |
| :--- | :--- | :--- |
| Supply voltage to ground | -0.3 to 5.5 | V |
| Input/output voltage to ground | -0.3 to VDD+0.3 | V |
| Operating temperature | -40 to +85 | ${ }^{\circ} \mathrm{C}$ |
| Storage temperature | -55 to +150 | ${ }^{\circ} \mathrm{C}$ |
| Soldering temperature/Time | $260^{\circ} \mathrm{C} / 10 \mathrm{Sec}$ |  |
| ESD immunity, Human Body Model/Machine Model | $\leq 1.5 \mathrm{kV} / 200 \mathrm{~V}$ |  |
| Latch-up immunity | $\leq 100 \mathrm{~mA}$ |  |

### 9.2 DC Characteristics

(VDD $=3 \mathrm{~V}, \mathrm{~T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$, unless otherwise noted)

| Symbol | Parameter | Test Conditions | Min. | Typ. | Max. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| VDD | Supply voltage | -40 to $+85{ }^{\circ} \mathrm{C}$ | 2.4 |  | 3.6 | V |
| IDD | Supply current | In operating mode |  | 1.5 |  | mA |
| IPD | Sleep mode supply current | In sleep mode, LVR enable |  | 2 |  | $\mu \mathrm{A}$ |
| VIH | Digital Input high voltage | PT1, RST_ | 0.7 |  |  | VDD |
| VIL | Digital Input low voltage | PT1, RST_ |  |  | 0.3 | VDD |
| VIHSH | Input Hys. High Voltage | Schmitt-trigger port |  | 0.45 |  | VDD |
| VIHSL | Input Hys. Low Voltage | Schmitt-trigger port |  | 0.20 |  | VDD |
| IOH | High Level Output Current | $\mathrm{VOH}=\mathrm{VDD}-0.3 \mathrm{~V}$ |  | 3 |  | mA |
| IOL | Low Level Output Current | $\mathrm{VOL}=0.3 \mathrm{~V}$ |  | 5 |  | mA |
| VSR | Voltage source switch resistor |  |  | 10 |  | $\Omega$ |
| VDDA | Analog Power for 3.6V Output (no load) for FS98002A and FS98O02C | VGG > 4V | 3.45 | 3.6 | 3.85 | V |
|  | Analog Power for $\mathrm{VDD}_{\text {(max. 3.6V) }}$ Output (no load) for FS98002B | $\mathrm{VDD} \leq 4 \mathrm{~V}$ |  |  | 3.6 | V |
|  | Analog Power for 2.5V Output ( no load) | VGG > 2.7V | 2.37 | 2.5 | 2.67 | V |
| KTCREF | VDDA temperature coefficient | $\mathrm{T}_{\mathrm{A}}=0 \sim 50^{\circ} \mathrm{C}$ |  | 100 |  | ppm/ ${ }^{\circ} \mathrm{C}$ |
| VLBAT | Low battery detection voltage | $\mathrm{VDD}=2.5 \mathrm{~V}$ <br> [ADOH,ADOM] is about 2710h | 2.2 | 2.4 | 2.6 | V |
| VLVR | Low voltage reset voltage |  |  | 1.65 |  | V |
| VLCD | LCD driver peak to peak voltage |  | 2.6 | 2.8 | 3.0 | V |
| FCK | Internal RC oscillator frequency for 1 MHz |  | 0.8 | 1.0 | 1.2 | MHz |
|  | Internal RC oscillator frequency for 4MHz |  | 3.0 | 4.0 | 4.8 | MHz |
| FWDT | Internal WDT Clock |  |  | 1.0 |  | kHz |
| VPP | Instruction Programmer input Voltage | PROEN = 0 | 11 | 12 | 13 | V |

### 9.3 ADC Characteristics

$\left(\mathrm{VDD}=3 \mathrm{~V}, \mathrm{~T}_{\mathrm{A}}=25^{\circ} \mathrm{C}\right.$, unless otherwise noted)

| Symbol | Parameter | Test Conditions | Min. | Typ. | Max. | Unit |
| :--- | :--- | :--- | :---: | :---: | :---: | :---: |
| VDIN | ADC differential input voltage range | To VSS | 1 |  | 2.2 | V |
| VRIN | ADC reference input voltage range | (VRH, VRL), ADC Gain $=1$ | 0.25 |  | 0.5 | V |
|  | Resolution |  |  | $\pm 15625$ |  | Counts |
|  | ADC linearity error | VRIN $=0.44 \mathrm{~V}$ for FRC $=1 \mathrm{MHz}$ | -0.1 | 0 | +0.1 | mV |

### 9.4 OPAMP Characteristics

(VDD $=3 \mathrm{~V}, \mathrm{~T}_{\mathrm{A}}=25^{\circ} \mathrm{C}$, unless otherwise noted)

| Symbol | Parameter | Test Conditions | Min. | Typ. | Max. | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | Input Offset |  |  | 1.5 |  | mV |
|  | Input Offset Voltage with Chopper | Rs<100 $\Omega$ |  | 20 |  | $\mu \mathrm{V}$ |
|  | Input Reference Noise | Rs=100 $\Omega, 0.1 \mathrm{~Hz} \sim 1 \mathrm{~Hz}$ |  | 1.0 |  | $\mu \mathrm{Vpp}$ |
|  | Input Reference Noise with Chopper | Rs=100 $\Omega, 0.1 \mathrm{~Hz} \sim 1 \mathrm{~Hz}$ |  | 0.5 |  | $\mu \mathrm{Vpp}$ |
|  | Input Bias Current |  |  | 10 | 30 | pA |
|  | Input Bias Current with Chopper |  |  | 100 | 300 | pA |
|  | Input Common Mode Range |  | 0.5 |  | 2.4 | V |
|  | Output Voltage Range |  | 0.5 |  | 2.4 | V |
|  | Chopper Clock Frequency | S_CHCK[1:0]=11 |  | 1k |  | Hz |
|  | Capacitor Load | , |  | 50 | 100 | pF |

## 10. Typical Performance Characteristics

1. $\operatorname{VDD}(\mathrm{V})$ vs. $\mathrm{VGG}(\mathrm{V})$

Temp $=25^{\circ} \mathrm{C}$, VGG Capacitor= $10 \mu \mathrm{~F}$, Charge pump Capacitor $\mathrm{CA}-\mathrm{CB}=10 \mu \mathrm{~F}$, VDDA and VS(no loading)

2. VDDA Load(mA) vs. VGG(V)

Temp $=25^{\circ} \mathrm{C}$, VGG Capacitor $=10 \mu \mathrm{~F}$, Charge pump Capacitor CA-CB=10 F


## 11. CPU Core

Figure 11-1 shows the CPU core block diagram used in FS98O02.


Figure 11-1: CPU Core Block Diagram

### 11.1 Program Memory Organization

The CPU has a 10-bit program counter capable of address up to $2 \mathrm{k} \times 16$ program memory space. The reset vector is at 0000h and the interrupt vector is at 0004 h .


Figure 11-2: Program Memory Origination

### 11.2 Data Memory Organization

The data memory is partitioned into three parts. The address $00 \mathrm{~h} \sim 07 \mathrm{~h}$ areas are system special registers, like indirect address, indirect address pointer, status register, working register, interrupt flag, interrupt control register. The address 08h~7Fh areas are peripheral special registers, like I/O ports, timer, ADC, signal
conditional network control register, LCD driver. The address $80 \mathrm{~h} \sim \mathrm{FFh}$ areas are general data memory.

Table 11-1: Data Memory Organization

| Address | Name | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Reset State | WDT Reset State | Details on page: |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 00h | INDO | Use contents of FSR0 to address data memory |  |  |  |  |  |  |  | unuu uuuu | uuuu uuuu | 16 |
| 02h | FSR0 | Indirect data memory, address point 0 |  |  |  |  |  |  |  | unuu uuuu | unuu uuuu | 16 |
| 04h | STATUS |  |  |  | PD | TO | DC | C | Z | ---0 Ouuu | ---u 1uuu | 16 |
| 05h | WORK | WORK register |  |  |  |  |  |  |  | uuuu uuuu | unuu uuuu | 16 |
| 06h | INTF |  |  |  | TMIF |  |  | ADIF | EOIF | ---0 --00 | ---0-00 | 25, 27, 28 |
| 07h | INTE | G |  |  | TMIE |  |  | ADIE | EOIE | 0--0 --00 | 0--0-00 | 25, 27, 28 |
| 08h~7Fh |  | Peripheral special registers |  |  |  |  |  |  |  |  |  | 18 |
| 80h~FFh |  | General data memory (128-byte SRAM) |  |  |  |  |  |  |  | unuu uuuu | unuu uuuu | , |

Note 1: "u" means unknown or unchanged. "-" means unimplemented, read as " 0 ".
Note 2: The "Reset State" indicates the registers state after external reset and low voltage reset.
Note 3: The "WDT Reset State" indicates the registers state after watchdog time out reset.

### 11.3 System Special Registers

System special registers are used by the CPU to control the operation of the device. Some registers are used for data memory access, some for logic judgment, and some for arithmetic, etc.

Table 11-2: System Special Registers

| Address | Name | $\begin{aligned} & \text { Bit } \\ & 7 \end{aligned}$ | $\begin{aligned} & \text { Bit } \\ & 6 \end{aligned}$ | $\begin{aligned} & \text { Bit } \\ & 5 \end{aligned}$ | Bit 4 | $\begin{aligned} & \text { Bit } \\ & 3 \end{aligned}$ | $\begin{aligned} & \text { Bit } \\ & 2 \end{aligned}$ | Bit 1 | Bit 0 | Reset State | WDT <br> State | Details page: |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 00h | IND0 | Use contents of FSR0 to address data memory |  |  |  |  |  |  |  | unuu uuuu | uuuu uuuu | 16 |
| 02h | FSR0 | Indirect data memory, address point 0 |  |  |  |  |  |  |  | unuu uuuu | unuu unuu | 16 |
| 04h | STATUS |  |  |  | PD | TO | DC | C | Z | ---0 Ouuu | ---u 1uuu | 16 |
| 05h | WORK | WORK register |  |  |  |  |  |  |  | uuuu uuuu | uuuu uuuu | 16 |
| 06h | INTF |  |  |  | TMIF |  |  | ADIF | E0IF | ---0 --00 | ---0 --00 | 25, 27, 28 |
| 07h | INTE | GIE |  |  | TMIE |  |  | ADIE | EOIE | 0--0 --00 | 0--0 --00 | 25, 27, 28 |

Note 1: "u" means unknown or unchanged. "-" means unimplemented, read as " 0 ".
Note 2: The "Reset State" indicates the registers state after external reset and low voltage reset.
Note 3: The "WDT Reset State" indicates the registers state after watchdog time out reset.

- INDO : Address 00h

The INDO registers at data memory address are not physical registers. Any instruction using the INDO register actually access the data pointed by the FSR0 register.

A simple program to clear data memory 80h-0BFh using indirect addressing is shown as Example 11-1.

|  | MOVLW | 080h |  |
| :--- | :--- | :--- | :--- |
|  | MOVWF | FSR0 |  |
| NEXT: | CLRF | IND0 | ; Clear the content of memory address pointed by FSR0 |
|  | INCFSZ | FSR0, 1 | ; FSR0 $=$ FSRO +1 , and judge if FSR0 $=0$ |
|  | GOTO | NEXT |  |

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- FSR0 : Address 02h

Indirect addressing pointers FSR0 correspond to IND0 respectively.

## - STATUS : Address 04h

The STATUS register contains the arithmetic status of the ALU. The function of each bit in STATUS register is described in Table 11-3.

Table 11-3: Status Register

| Bit | Symbol | Description |
| :--- | :--- | :--- |
| $7 \sim 5$ | - | No use. |
| 4 | PD | Power down flag. <br> 1: After power on reset or cleared by writing 0 (which shuts off oscillator clock, thus neither of the MCU <br> clock or operation will be in conduct). <br> 0: By execution of the SLEEP instruction, but not the HALT instruction (which only turns off the MCU <br> clock). |
| 3 | - | No use. |
| 2 | DC | Digit carry flag (ADDWF, SUBWF instructions) <br> 1: A carry-out from the 4th low order bit of the result occurred. <br> 0: No carry-out from the 4th low order bit of the result. |
| 1 | C | Carry flag (~Borrow) <br> 0 | | Zero flag |
| :--- |
| 1: The result of an arithmetic or logic operation is zero. |
| 0: The result of an arithmetic or logic operation is not zero. |

## - WORK : Address 05h

WORK register is used to store temporary data for arithmetic, data moving, etc.

## - INTF, INTE: Address 06h, 07h

The interrupt enable register (INTE) records individual interrupt request. When some interrupt event occurs and related interrupt enable bit $=1$, the related interrupt flag in interrupt flag register (INTF) will be set. The global interrupt enable bit (GIE) will enable CPU interrupt procedure. When $\mathrm{GIE}=1$ and any interrupt flag is set, CPU interrupt procedure would be executed. CPU interrupt procedure executes GIE reset and CALL 0004h.

When interrupt signal happened within instruction duty cycle, the CPU must wait and till instruction duty cycle end of this program then produce an "Interrupt Flag" before go into next step.

This Example 11-2 program is specially mentioned here for halt and sleep mode.

| MAIN: |  |  |
| :--- | :--- | :--- |
|  | HALT |  |
|  | NOP |  |
| GOTO | MAIN |  |
| MAIN_SLEEP: |  |  |
|  | CLRF | INTF |
| SLEEP |  |  |
|  |  |  |
|  |  |  |
|  |  |  |
|  |  |  |

Example 11-2: Halt and Sleep Mode Example

Note. Please make sure all interrupt flags are cleared before running SLEEP; "NOP" command must follow HALT and SLEEP commands.

The INTF register contains the status of every interrupt event. The function of each bit in INTF register is described in Table 11-4.

Table 11-4: INTF Register

| Bit | Symbol | Description |
| :--- | :--- | :--- |
| $7 \sim 5$ | - | No use. |
| 4 | TMIF | 8-bit timer Interrupt flag. |
| $3 \sim 2$ | - | No use. |
| 1 | ADIF | Analog to digital converter Interrupt flag. |
| 0 | EOIF | PT1<0> external interrupt flag. |

The INTE register defines if the CPU will accept related interrupt event. The function of each bit in INTE register is described in Table 11-5.

Table 11-5: INTE Register

| Bit | Symbol | Description |
| :--- | :--- | :--- |
| 7 | GIE | Global interrupt enable bit. |
| $6 \sim 5$ | - | No use. |
| 4 | TMIE | 8-bit timer Interrupt enable bit. |
| $3 \sim 2$ | - | No use. |
| 1 | ADIE | Analog to digital converter Interrupt enable bit. |
| 0 | EOIE | PT1<0> external interrupt enable bit. |

### 11.4 Peripheral Special Registers

The peripheral special registers are used to control I/O ports, timer, ADC, signal conditional network, LCD driver, and others.

Table 11-6: Peripheral Special Registers

| Address | Name | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Reset State | WDT <br> Reset <br> State | Details on page |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 09h | PCK |  |  |  |  | S_CHC | [1:0] | S_BEEP | EN_OSC4M | ---- 0000 | ---- 0000 | 24, 25 |
| OAh | EADRH | PAR[15:8] |  |  |  |  |  |  |  | unuu unuu | uuuu unuu | 35 |
| OBh | EADRL | PAR[7:0] |  |  |  |  |  |  |  | unuu unuu | unuu unuu | 35 |
| OCh | EDAH | EDATA[15:8] |  |  |  |  |  |  |  | unuu unuu | unuu unuu | 35 |
| ODh | TMOUT | TMOUT [7:0] |  |  |  |  |  |  |  | 00000000 | 00000000 | 25, 27 |
| OEh | TMCON | TMRST | WDTEN | WTS [1:0] |  | TMEN | INS [2:0] |  |  | 10000000 | 1u00 0000 | 25, 27 |
| OFh | TMMOD | TMMOD [7:0] |  |  |  |  |  |  |  | 00000000 | 00000000 | 27 |
| 10h | ADOH | ADO [23:16] |  |  |  |  |  |  |  | 00000000 | 00000000 | 29 |
| 11h | ADOM | ADO [15:8] |  |  |  |  |  |  |  | 00000000 | 00000000 | 29 |


| Address | Name | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Reset State | WDT Reset State | Details on page: |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 12h | ADOL | ADO [7:0] |  |  |  |  |  |  |  | 00000000 | 00000000 | 29 |
| 13h | ADCON |  |  |  |  | ADRST | ADM [2:0] |  |  | ---- 0000 | ---- 0000 | 29 |
| 1Fh | SVD |  |  |  |  |  |  | $\square$ | SVD | ---- ---1 | ---- ---1 | 19, 22 |
| 20h | PT1 | PT1 [7:0] |  |  |  |  |  |  |  | unuu unuu | unuu unuu | 25 |
| 21 h | PT1EN | PT1EN[7:4] |  |  |  | - |  | $\square$ | , | 0000 ---- | 0000 ---- | 25 |
| 22h | PT1PU | PT1PU [7:0] |  |  |  |  |  |  |  | 00000000 | 00000000 | 25 |
| 23h | PT1MR | BPE | EN_LBD |  |  |  |  | EOM [1:0] |  | 00-- --00 | 00-- --00 | 25 |
| 24h | PT2 |  |  |  |  | PT2 [3:0] |  |  |  | ----- uиuu | ----- uuuu | 26 |
| 25h | PT2EN |  |  |  |  | PT2EN [3:0] |  |  |  | ---- 0000 | ---- 0000 | 26 |
| 26 h | PT2PU |  |  |  |  | PT2PU [3:0] |  |  |  | ---- 0000 | ---- 0000 | 26 |
| 2Ah | NETB |  | SINL[1:0] |  | SINH [1:0] |  | SFT [2] |  | SFT [0] | -000 00-0 | -000 00-0 | 30 |
| 2Ch | NETD | EPMAT | SVRH[0] | SVRL[1:0] |  | ERV | EPBLK | SLVD | SVR | u000 uu00 | u000 uu00 | 30, 35 |
| 2Eh | NETF | EN_PUMP | EN_VS | EN_LC | LCDEN | EN_VGG3 | EN_VDDAPU | EN_VDDA25 | EN_VDDA | 00000100 | 00000100 | 19, 36 |
| 2Fh | NETG | - | , |  |  | ADG [1:0] |  | ADEN | AZ | ---- 0000 | ---- 0000 | 29 |
| 33h | NETK | SILB[2:0] |  |  | EN_LB | OP1EN | SOP1P[1:0] |  | SOP1N | 00000000 | 00000000 | 22, 34 |
| 40h | LCD1 | SEG1 [3:0] |  |  |  | SEG0 [3:0] |  |  |  | unuu uuuu | uиuu unuu | 36 |
| 41h | LCD2 | SEG3 [3:0 |  |  |  | SEG2 [3:0] |  |  |  | unuu unuu | unuu unuu | 36 |
| 42h | LCD3 | SEG5 [3:0] |  |  |  | SEG4 [3:0] |  |  |  | unuu uuuu | unuu uuuu | 36 |
| 43h | LCD4 | SEG7 [3:0] |  |  |  | SEG6 [3:0] |  |  |  | unuu unuu | unuu unuu | 36 |
| 44h | LCD5 | SEG9 [3:0] |  |  |  | SEG8 [3:0] |  |  |  | unuu unuu | unuu unuu | 36 |
| 45h | LCD6 | SEG11 [3:0] |  |  |  | SEG10 [3:0] |  |  |  | unuu unuu | unuu unuu | 36 |

Note 1: "u" means unknown or unchanged. "-" means unimplemented, read as " 0 ".
Note 2: The "Reset State" indicates the registers state after external reset and low voltage reset.
Note 3: The "WDT Reset State" indicates the registers state after watchdog time out reset.

## 12. Power System

There're some important power management blocks in FS98O02, such as voltage doublers, voltage regulator, analog bias circuit, analog common voltage generator, etc. The power system related registers are in Table 12-1.

Table 12-1: Power System Related Registers

| Address | Name | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Reset State | WDT <br> Reset State |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| 1Fh | SVD |  |  |  |  |  |  |  | SVD | ------1 | ------1 |
| 2Eh | NETF | EN_PUMP | EN_VS | EN_LCDB | LCDEN | EN_VGG3 | EN_VDDAPU | EN_VDDA25 | EN_VDDA | 00000100 | 00000100 |
| 33h | NETK | SILB[2:0] |  | EN_LB |  |  |  |  | 00000000 | 00000000 |  |

### 12.1 Voltage Doubler

The voltage doubler is used to generate double voltage of VDDP or 3V VGG output. The doubled voltage of VDDP is used for regulator power supply and the 3V VGG output can be used for LCD power supply. When voltage doubler is turned off ( EN_PUMP = 0), the default VGG is shorted to VDDP and this VGG output can be used for LCD power supply. During the LCD application, if the VDDP voltage is reduced, the brightness of the LCD may be darkened. User can turn on voltage doubler by EN_PUMP after setting the EN_VGG3 to pump the VGG voltage to 3 V to supply LCD circuit. It can lighten LCD and hold the LCD brightness.

Note. Before turn on the 3 V VGG application, please turn on the Regulator (EN_VDDA=1) first. On the other hand, to turn on the double voltage of VDDP application, the Regulator can be turn off.


Figure 12-1: Voltage Doubler
When EN_PUMP = 1, voltage doubler is enabled. The VGG voltage is about two times of VDDP when EN_VGG3 $=0$ (default) or about 3 V when EN_VGG3 $=1$ and EN_VDDA $=1$. When EN_PUMP $=0$, you can input a voltage as voltage regulator power supply.

Voltage doubler operation frequency is selected by S_PCK. The details are described in "Voltage Doubler Operation Frequency" on page 28.

Typical value for C 1 or C 2 is $1 \mu \mathrm{~F} \sim 10 \mu \mathrm{~F}$. For large load current, larger capacitors should be used to reduce the output voltage ripple. If a polarity capacitor is used for C1, the CB pin should be connected to the negative terminal of the capacitor, and the CA pin to the positive terminal.

### 12.2 The FS98002A and FS98002C Voltage Regulator

The voltage regulator is used to regulate the doubled voltage of VDDP from VGG to analog power supply, VDDA. VDDA is the power supply voltage for analog circuit and LCD driver.


Figure 12-2: The FS98002A and FS98002C Voltage Regulator
When EN_VDDA = 1, the voltage regulator will be enabled and the VDDA is about 3.6 V when EN_VDDA25 $=0$ (default) or about 2.5 V when EN_VDDA25 $=1$. Otherwise VDDA can be used as external
regulated power supply input after turn off EN_VDDA and EN_VDDAPU. The default of the EN_VDDAPU is turned on to enable the pull-high circuit to pull high the VDDA to VDD when regulator is turned off. After turning on the regulator, the pull-high circuit will be turned off automatically until the regulator being turned off. If turn off the EN_VDDAPU, the VDDA pin will be floating when regulator is turned off.

The typical capacitance for C 3 is $1 \mu \mathrm{~F} \sim 10 \mu \mathrm{~F}$. For large load current, large capacitor should be used to increase the output voltage stability.

Note. Except the VDDA pin is used for input, the EN_VDDAPU should be turn on (default) .

### 12.3 The FS98O02B Voltage Regulator

The voltage regulator is used to regulate the voltage of the VDD to analog power supply, VDDA. VDDA is the power supply voltage for analog circuit and LCD driver.


When EN_VDDA = 1, the voltage regulator will be enabled and the VDDA is about $\mathrm{VDD}_{(\max .3 .6 \mathrm{~V})}$ when EN_VDDA25 $=0$ (default) or about 2.5 V when EN_VDDA25 $=1$. Otherwise VDDA can be used as external regulated power supply input after turn off EN_VDDA and EN_VDDAPU. The default of the EN_VDDAPU is turned on to enable the pull-high circuit to pull high the VDDA to VDD when regulator is turned off. After turning on the regulator, the pull-high circuit will be turned off automatically until the regulator being turned off. If turn off the EN_VDDAPU, the VDDA pin will be floating when regulator is turned off.

The typical capacitance for C 3 is $1 \mu \mathrm{~F} \sim 10 \mu \mathrm{~F}$. For large load current, large capacitor should be used to increase the output voltage stability.
Note. Except the VDDA pin is used for input, the EN_VDDAPU should be turn on (default) .

### 12.4 Analog Bias Circuit



Figure 12-4: Analog Bias Circuit

Before enabling the analog block, EN_VDDA must be set. When the internal voltage doubler is used, a 10 nF capacitor must be connected between pin VB and VSS for reducing voltage doubler's noise.

### 12.5 Analog Common Voltage Generator

Analog common voltage generator is used to generate the analog common voltage, AGND.


Figure 12-5: Analog Common Voltage Generator

When EN_VDDA = 1, analog common voltage generator is enabled. AGND voltage is about $1 / 2$ VDDA.

### 12.6 Low Battery Detector

Low battery detector is used for VDD low voltage detection. FS98002 embeds a voltage divider which can generate $2.3 \mathrm{~V}, ~ 2.4 \mathrm{~V}, ~ 2.5 \mathrm{~V}, ~ 2.6 \mathrm{~V}$ and 2.7 V . A multiplexer is used to connect the voltage divides to component input. The multiplexer's output is compares with 1.2 V . The Control register flags are SILB[2:0] and the EN_LB.

The output flag is SVD which is for read only. Please see Figure 12-6


Figure 12-6: Low Battery Detector Block

Table 12-2: Low battery detector voltage detection selection table

| SILB[2:0] | Detection Voltage | The SVD=1 Condition |
| :--- | :--- | :--- |
| 000 | $\mathrm{VDD}=2.3 \mathrm{~V} \pm 100 \mathrm{mV}$ | $\mathrm{VDD} \geq 2.2 \mathrm{~V}$ |
| 001 | $\mathrm{VDD}=2.4 \mathrm{~V} \pm 100 \mathrm{mV}$ | $\mathrm{VDD} \geq 2.3 \mathrm{~V}$ |
| 010 | $\mathrm{VDD}=2.5 \mathrm{~V} \pm 100 \mathrm{mV}$ | $\mathrm{VDD} \geq 2.4 \mathrm{~V}$ |
| 011 | $\mathrm{VDD}=2.6 \mathrm{~V} \pm 100 \mathrm{mV}$ | $\mathrm{VDD} \geq 2.5 \mathrm{~V}$ |
| 100 | $\mathrm{VDD}=2.7 \mathrm{~V} \pm 100 \mathrm{mV}$ | $\mathrm{VDD} \geq 2.6 \mathrm{~V}$ |
| 101 | $\mathrm{PT} 1.6=1.2 \mathrm{~V} \pm 100 \mathrm{mV}$ | $\mathrm{PT} 1.6 \geq 1.1 \mathrm{~V}$ |
| 11 X | NA | NA |

### 12.7 LCD Bias Circuit

V3, V2, V1 in Figure 12-7 are the output voltages of the LCD bias circuit. The voltages to VSS are about $V_{\text {RLCDin }}, 2 / 3 * V_{\text {RLCDin }}$ and $1 / 3 * V_{\text {RLCDin }}$, and the voltages are used in LCD driver.



Figure 12-7: LCD Bias Circuit

When EN_LCDB = 1 , the LCD bias circuit is enabled. When EN_LCDB $=0$, the LCD bias circuit is disabled, and then the LCD driver will not function.

## 13. Clock System

The clock system offers several clocks to some important blocks in FS98002, such as CPU clock, ADC sample frequency, beeper clock, voltage doubler operating frequency, etc. Only with the clock signals from the clock system, the FS98O02 can work normally.
Table 13-1: Clock System Related Registers

| Address | Name | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Reset State | WDT Reset State |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| 09h | PCK |  |  |  |  | S_CHCK[1:0] | S_BEEP | EN_OSC4M | ---0000 | ---0000 |  |



Figure 13-1: Internal Clock for CPU and Other Blocks

### 13.1 Oscillator State

MCK is the heart of the clock system. Almost all clock signals are derived from the MCK. If we stop MCK, many clock signals will be stopped. We may use "sleep" instruction to disable MCK as described in Table 13-2.

When EN_OSC4M = 0 (default), the MCK is about 1 MHz . When EN_OSC4M = 1 , the MCK is about 4 MHz .

Table 13-2: How to Use MCK

| Sleep | MCK |
| :--- | :--- |
| 1 | Disable |
| 0 | Enable |

### 13.2 CPU Instruction Cycle

The CPU in FS98O02 has only one mode of instruction cycle. That is MCK/4 ( $\sim 4$ us for 1 MHz MCK or $\sim 1$ us for 4 MHz MCK).

### 13.3 ADC Sample Frequency

ADC sample frequency decides the sampling rate of the input signal. The sampling rate of ADC in FS98002 is fixed to MCK/25.

Table 13-3: ADC Sampling Frequency

| ADC Sample Frequency (ADCF) |
| :--- |
| $\mathrm{MCK} / 25(\sim 40 \mathrm{kHz}$ for 1 MHz MCK or $\sim \mathbf{1 6 0 k H z}$ for 4 MHz MCK$)$ |

### 13.4 Beeper Clock

We may set beeper clock by select the proper value of S_BEEP as in Table 13-4.

FS98002

| S_BEEP | Beeper Clock |
| :--- | :--- |
| 1 | MCK/500 for $1 \mathrm{MHz} \mathrm{MCK} \mathrm{or} \mathrm{MCK/2000} \mathrm{for} 4 \mathrm{MHz} \mathrm{MCK} \mathrm{( } \mathrm{\sim 2kHz)}$ |
| 0 | MCK/312 for 1 MHz MCK or MCK/1248 for $4 \mathrm{MHz} \mathrm{MCK}(\sim 3.2 \mathrm{kHz})$ |

### 13.5 Voltage Doubler Operation Frequency

The voltage doubler operation frequency is related to the load capability. There is an internal S_PCK signal deciding the voltage doubler operating frequency.

Table 13-5: Voltage Doubler Operation Frequency

| S_PCK | Voltage Doubler Operation Frequency |
| :--- | :--- |
| 1 | MCK/50 for 1MHz MCK or MCK/200 for $4 \mathrm{MHz} \mathrm{MCK}(\sim 20 \mathrm{kHz})$ |

### 13.6 Timer and LCD Module Input Clock

The timer and LCD module input clock in FS98002A and FS98002B is MCK/2000 for 1 MHz MCK or MCK/8000 for 4 MHz MCK. The timer and LCD module input clock in FS98002C is MCK/25 for 1 MHz or 4 MHz MCK as described in Table 13-6.

Table 13-6: Timer and LCD Module Input Clock

## Timer and LCD Module Input Clock TMCLK

MCK/2000 for 1MHz MCK or MCK/8000 for 4MHz MCK (~500Hz) for FS98002A and FS98002B
MCK/25 for 1 MHz or 4 MHz MCK for FS98O02C

### 13.7 OPAM Chopper Input Clock

The OPAM chopper input clock in FS98002 is selected by internal S_CH1CK as described in Table 13-7.

Table 13-7: OPAMP chopper Input Clock

| S_CH1CK[1:0] | OPAMP chopper mode (input operation) |
| :---: | :--- |
| 00 | +Offset |
| 01 | - Offset |
| 10 | MCK/500 for 1 MHz MCK or MCK/2000 for 4MHz MCK chopper frequency |
| 11 | MCK/1000 for 1 MHz MCK or MCK/4000 for 4MHz MCK chopper frequency |

## 14. I/O Port

We may set the I/O port to be input port or output port in FS98O02 for our applications. We may also set PT1 [7] as buzzer output to drive the external buzzer to generate sounds. The buzzer's beeper frequency is show in Table 13-4.

Table 14-1: I/O Port Related Registers

| Address | Name | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Reset State | WDT Reset State |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 06h | INTF |  |  |  |  |  |  |  | EOIF | ---0 --00 | ---0 --00 |
| 07h | INTE | GIE |  |  |  |  |  |  | EOIE | 0--0 --00 | 0--0 --00 |
| 20h | PT1 | PT1 [7:0] |  |  |  |  |  |  |  | uuuu uuuu | uuuu uuuu |
| 21h | PT1EN | PT1EN [7:4] |  |  |  |  |  |  |  | 0000 ---- | 0000 ---- |
| 22h | PT1PU | PT1PU [7:0] |  |  |  |  |  |  |  | 00000000 | 00000000 |
| 23h | PT1MR | BPE | - |  |  |  |  | EOM |  | 0----00 | 0----00 |
| 24h | PT2 | - |  |  |  | PT2 [3:0] |  |  |  | ---- uuuu | ---- uuuu |


| Address | Name | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Reset State | WDT Reset State |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| 25 h | PT2EN |  |  |  |  | PT2EN $[3: 0]$ |  | ---0000 | ---0000 |  |  |
| 26 h | PT2PU |  |  |  |  | PT2PU $[3: 0]$ | ---0000 | ---0000 |  |  |  |

### 14.1 PT1

PT1 is 4-bit Input port and 4-bit I/O port with pull-up resistor enable control.
PT1 [ N$]$ is an input port when PT1EN $[\mathrm{N}]=$ " 0 "; PT1 [ N$]$ is an output port when PT1EN $[\mathrm{N}]=$ " 1 ". When PT1EN [7] = " 1 " and BPE = " 1 ", PT1 [7] is used as the buzzer output.

When PT1PU $[\mathrm{N}]=$ " 0 ", PT1 $[\mathrm{N}]$ has no pull-up resistor; When PT1PU $[\mathrm{N}]=$ " 1 ", PT1 $[\mathrm{N}]$ has a pull-up resistor.

PT1 [0] can be used as an external interrupt source. Interrupt mode of PT1 [0] is controlled by EOM [1:0]. When EOM [1:0] = "00", PT1 [0] is for negative edge trigger interrupt input; and "01" for positive edge. " 10 " \& " 11 " for interrupt during other time.

If VPP $=12 \mathrm{~V}, \mathrm{PT} 1[1]$ use for judge if SPI or instruction programmer EPROM, if PT1[1] connect to VSS then used instruction programmer EPROM function. If PT1[1] connect to VDD then used SPI programmer EPROM.

If VPP $=0 \sim 3.6 \mathrm{~V}, \mathrm{PT} 1[1]$ is a input port.
PT1 [ N ] has Schmitt-trigger input.


Figure 14-1: the Block Diagram of Port 1

### 14.2 PT2

PT2 is 4-bit I/O port with pull-up resistor enable control. PT2 [ N$]$ is an input port when PT2EN $[\mathrm{N}]=$ " 0 "; PT2 $[\mathrm{N}]$ is an output port when PT2EN $[\mathrm{N}]=$ " 1 ". When PT2PU $[\mathrm{N}]=$ " 0 ", PT2 $[\mathrm{N}]$ has no pull-up resistor; When PT2PU $[\mathrm{N}]=$ "1", PT2 [ N$]$ has a pull-up resistor. PT2 [ N$]$ has Schmitt-trigger input.


Figure 14-2: the Block Diagram of Port 2

## 15. 8-bit Timer

The 8-bit timer in FS98O02 is usually used for timer interrupt in applications. We may have a periodic internal interrupt to do some periodic procedure in CPU by using the 8 -bit timer properly.

Table 15-1: 8-bit Timer Related Registers

| Address | Name | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Reset State | WDT Reset State |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 06h | INTF | - |  |  | TMIF |  |  |  |  | ---0 --00 | ---0 --00 |
| 07h | INTE | GIE |  |  | TMIE |  |  |  |  | 0--0 --00 | 0--0 --00 |
| 0Dh | TMOUT | TMOUT [7:0] |  |  |  |  |  |  |  | 00000000 | 00000000 |
| 0Eh | TMCON | TMRST |  |  |  | TMEN | INS |  |  | 10000000 | 14000000 |
| 0Fh | TMMOD | TMMOD [7:0] |  |  |  |  |  |  |  | 00000000 | 00000000 |



Figure 15-1: The 8-bit Timer Block Diagram
After writing a " 0 " to bit 7 of address 0 Eh, the CPU will send a low pulse to TMRST and reset the 8 -bit counter and TMCLK divider. And then the bit 7 of address 0 EH will be set to " 1 ".

When TMEN $=1$, the register value of TMMOD is zero or not that controls which one of the 8 -bit counter or TMCLK divider is enabled. When TMMOD $=8$ 'h00, the 8 -bit counter is enabled and the TMCLK divider is disabled. Otherwise, if the TMMOD has been set not zero value, the TMCLK divider is enabled and 8 -bit counter is disabled. When TMEN $=0$, the 8 -bit counter and TMCLK divider are all stopped.

INS [2:0] and TMMOD select timer interrupt source. When TMMOD $=8$ 'h00, the selection codes are described in Table 15-3. Otherwise the selection codes are described in Table 15-4.

Table 15-2: The Timer Input Clock TMCLK

| Product | The TMCLK for Int. OSC=1MHz | The TMCLK for Int. OSC=4MHz |
| :--- | :---: | :---: |
| FS98O02A / FS98O02B | 500 Hz | 500 Hz |
| FS98O02C | 40 KHz | 160 KHz |

Table 15-3: Setting Timer Interrupt Source by INS

| INS [2:0] | Timer Interrupt Source for FS98002A / B | Timer Interrupt Source for FS98O02C |
| :--- | :--- | :--- |
| 000 | TMOUT [0] (TMCLK/8 $: \sim 62.5 \mathrm{~Hz})$ | TMOUT [0] $\quad($ TMCLK/8 $: 5 \mathrm{KHz}$ or 20KHz $)$ |
| 001 | TMOUT [1] (TMCLK/16 $: \sim 31.25 \mathrm{~Hz})$ | TMOUT [1] (TMCLK/16 $: 2.5 \mathrm{KHz}$ or 10KHz ) |
| 010 | TMOUT [2] (TMCLK/32 $: \sim 15.625 \mathrm{~Hz})$ | TMOUT [2] (TMCLK/32 $: 1.25 \mathrm{KHz}$ or 5KHz ) |


| 011 | TMOUT [3] | ( TMCLK/64 : $\sim 7.813 \mathrm{~Hz}$ ) | TMOUT [3] | ( TMCLK/64: 625 Hz or 2.5 KHz ) |
| :---: | :---: | :---: | :---: | :---: |
| 100 | TMOUT [4] | ( TMCLK/128 : $\sim 3.91 \mathrm{~Hz}$ ) | TMOUT [4] | ( TMCLK/128:312.5Hz or 1.25 KHz ) |
| 101 | TMOUT [5] | ( TMCLK/256 : $\sim 1.953 \mathrm{~Hz}$ ) | TMOUT [5] | ( TMCLK/256: 156.25 Hz or 625 Hz ) |
| 110 | TMOUT [6] | ( TMCLK/512 : $\sim 0.977 \mathrm{~Hz}$ ) | TMOUT [6] | ( TMCLK/512: 78.125 Hz or 312.5 Hz ) |
| 111 | TMOUT [7] | ( TMCLK/1024 : $\sim 0.488 \mathrm{~Hz}$ ) | TMOUT [7] | (TMCLK/1024:39.0625Hz or 156.25 Hz ) |

Table 15-4: Setting Timer Interrupt Source by TMMOD

| TMMOD (0x0F ) | Timer Interrupt Source |
| :--- | :--- |
| 8'h00 | The Timer Interrupt Source is selected by INS[2:0] |
| 8'h01 | TMCLK / 2 |
| 8'h02 | TMCLK / 3 |
| 8'h03 | TMCLK / 4 |
|  |  |
| 8'hFE |  |
| 8'hFF | TMCLK / 255 |

When TMMOD $=8$ 'h00 and $\operatorname{INS}[2: 0]=33^{\prime} \mathrm{b} 000$, the timer interrupt function is equal to TMMOD $=8^{\prime} \mathrm{h} 07$ because of the timer interrupt sources are TMCLK/8. In the same manner, the timer interrupt source of the TMMOD $=8^{\prime} \mathrm{h} 00$ and $\operatorname{INS}[2: 0]=3^{\prime} \mathrm{b} 001$ is equal to $\mathrm{TMMOD}=8^{\prime} \mathrm{h} 0 \mathrm{~F}$. And the timer interrupt source of the TMMOD $=8$ 'h00 and $\operatorname{INS}[2: 0]=3^{\prime} \mathrm{b} 101$ is equal to TMMOD $=8$ 'hFF, etc.

TMOUT [7:0] is the output of the 8 -bit counter. It is a read-only register.

## 16. ADC

The ADC in the IC is a $\triangle \Sigma$ ADC with fully differential inputs and fully differential reference voltage inputs. Its maximum digital output code is $\pm 15625$.

The conversion equation is: Dout $=15625$ * $\mathrm{G}^{*}(\mathrm{VIH}-\mathrm{VIL}+\mathrm{Vio}) /(\mathrm{VRH}-\mathrm{VRL}+\mathrm{Vro})$. VIH is the ADC positive input voltage. VIL is the ADC negative input voltage. Vio is the ADC input offset voltage. VRH is the ADC positive reference voltage. VRL is the ADC negative reference voltage. Vro is the ADC reference offset voltage. And (VRH - VRL + Vro) > 0 . When G * (VIH - VIL + Vio) / (VRH - VRL + Vro) $\geqq$ 1, Dout $=15625$. When G * $(\mathrm{VIH}-\mathrm{VIL}+\mathrm{Vio}) /(\mathrm{VRH}-\mathrm{VRL}+\mathrm{Vro}) \leqq-1$, Dout $=-15625$.

### 16.1 ADC Digital Output Code Format

ADO [23:8] is the ADC digital output code. The digital output code is in 2's complement format, and the ADO [23], the most significant bit (MSB) of ADO represents the sign of the code. For example, if $A D O$ [23:8] = E2F7h, then Dout $=-(\operatorname{not}(E 2 F 7 h)+1)=-7433$.

### 16.2 ADC Linear Range

The $\Delta \Sigma$ ADC is close to saturation state when $\mathrm{G}^{*}(\mathrm{VIH}-\mathrm{VIL}+\mathrm{Vio}) /(\mathrm{VRH}-\mathrm{VRL}+\mathrm{Vro})$ is close to $\pm 1$. The


### 16.3 ADC Control Register

There are some ADC control related registers in FS98O02. There will be some more detail descriptions about using the ADC control register to get the proper ADC operation in users' applications.

Table 16-1: ADC Control Related Registers

| Address | Name | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Reset State | WDT Reset State |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 06h | INTF | - |  |  |  |  |  | ADIF |  | ---0 --00 | ---0 --00 |
| 07h | INTE | GIE |  |  |  |  |  | ADIE |  | 0--0 --00 | 0--0 --00 |
| 10h | ADOH | ADO [23:16] |  |  |  |  |  |  |  | 00000000 | 00000000 |
| 11h | ADOM | ADO [15:8] |  |  |  |  |  |  |  | 00000000 | 00000000 |
| 12h | ADOL | ADO [7:0] |  |  |  |  |  |  |  | 00000000 | 00000000 |
| 13h | ADCON | , |  |  | $3$ | ADRS | ADM [2:0] |  |  | ---- 0000 | ---- 0000 |
| 2Fh | NETG | $\square$ |  |  |  | ADG |  | ADEN | AZ | ---- 0000 | ---- 0000 |



Figure 16-1: the Block Diagram of ADC

The ADC contains a $\Delta \Sigma$ modulator and a digital comb filter. When ADEN $=1$, the $\Delta \Sigma$ modulator (ADC) will be enabled. When ADEN $=0$, the $\Delta \Sigma$ modulator (ADC) will be disabled. When ADRST $=1$, the digital comb filter will be enabled. When ADRST $=0$, the digital comb filter will be reset.

### 16.3.1 ADC Output Rate and Settling Time

The $\triangle \Sigma$ ADC is generally an over-sampling ADC. The ADC's each digital output code is the result of sampling the input signal N times and processed by DSP. The ADC's sampling frequency is decided by ADCF.

ADM decides when to send out a 16 -bit digital code after sampling $N$ times, and raises an interrupt signal every time the ADC produces a digital output code. In fact, the ADC's each digital output code is the result from the previous $2^{*} \mathrm{~N}$ times sampling results. If any of the ADC's input, reference voltage, ADG or AZ is switched, the first two output codes are normally not stable, and the third output code and later codes are stable for calculation.

The ADC's output rate is selected by ADM $[2: 0]$ as described in Table 16-2.

Table 16-2: ADC Output Rate

| ADM [2:0] | ADC Output Rate |
| :--- | :--- |
| 000 | ADCF/125 $(\sim 320 \mathrm{~Hz}$ for 1 MHz MCK or $\sim 1.28 \mathrm{kHz}$ for 4 MHz MCK$)$ |
| 001 | ADCF/250 $(\sim 160 \mathrm{~Hz}$ for 1 MHz MCK or $\sim 640 \mathrm{~Hz}$ for $4 \mathrm{MHz} \mathrm{MCK)}$ |


| 010 | ADCF/500 $(\sim 80 \mathrm{~Hz}$ for 1 MHz MCK or $\sim 320 \mathrm{~Hz}$ for 4 MHz MCK $)$ |
| :--- | :--- |
| 011 | ADCF/1000 $(\sim 40 \mathrm{~Hz}$ for $1 \mathrm{MHz} \mathrm{MCK} \mathrm{or} \sim 160 \mathrm{~Hz}$ for 4 MHz MCK$)$ |
| 100 | ADCF/2000 $(\sim 20 \mathrm{~Hz}$ for $1 \mathrm{MHz} \mathrm{MCK} \mathrm{or} \sim 80 \mathrm{~Hz}$ for $4 \mathrm{MHz} \mathrm{MCK)}$ |
| 101 | ADCF/4000 $(\sim 10 \mathrm{~Hz}$ for $1 \mathrm{MHz} \mathrm{MCK} \mathrm{or} \sim 40 \mathrm{~Hz}$ for $4 \mathrm{MHz} \mathrm{MCK)}$ |
| 110 | ADCF/8000 $(\sim 5 \mathrm{~Hz}$ for 1 MHz MCK or $\sim 20 \mathrm{~Hz}$ for $4 \mathrm{MHz} \mathrm{MCK)}$ |

### 16.3.2 ADC Input Offset

The ADC input offset voltage, Vio drifts with temperature and common mode voltage at the inputs. When $\mathrm{AZ}=0$, the $\Delta \Sigma$ modulator's differential inputs are (VIH, VIL); when $\mathrm{AZ}=1$, the $\Delta \Sigma$ modulator's differential inputs are (VIL, VIL). We can set AZ $=1$ to measure the ADC's offset.

When the drifting is slow, we may set $A Z=1$, and get Doff $=15625$ * $G *(V i o) /(V R H-V R L+V r o)$. When measuring input signal, Doff should be deducted.

### 16.3.3 ADC Gain

The ADC digital output code deducted by Doff is the ADC Gain. The ADC gain does not change as VDD changes. The suggested values for common mode voltages at ADC input and reference voltage are $1 \mathrm{~V} \sim 2 \mathrm{~V}$ respect to VSS. ADG [1:0] can set ADC's input gain: 00 for $2 / 3,01$ for 1,10 for 2 , and 11 for $7 / 3$.

### 16.3.4 ADC Resolution

The ADC resolution is mainly decided by ADM [2:0] (ADC output rate) and reference voltage, and the test results in FSC are as below for users' reference. When we set $(\mathrm{VRH}, \mathrm{VRL})=0.4 \mathrm{~V},(\mathrm{VIH}, \mathrm{VIL})=0.2 \mathrm{~V}, \mathrm{VRL}=\mathrm{VIL}$ = AGND, $\mathrm{G}=1$, and record ADO [23:8], we get the result in Table 16-3
. When we set $(\mathrm{VRH}, \mathrm{VRL})=\mathrm{VR},(\mathrm{VIH}, \mathrm{VIL})=1 / 2 * V R, V R L=\mathrm{VIL}=\mathrm{AGND}, \mathrm{G}=1, \mathrm{ADM}[2: 0]=101$, and record ADO [15:0], we get the result in Table 16-4.

Table 16-3

| ADM [2:0] | $\mathbf{0 0 0}$ | $\mathbf{0 0 1}$ | $\mathbf{0 1 0}$ | $\mathbf{0 1 1}$ | $\mathbf{1 0 0}$ | $\mathbf{1 0 1}$ | $\mathbf{1 1 0}$ |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| Rolling counts | 10 | 6 | 4 | 3 | 3 | 2 | 1 |

Table 16-4

| VR (V) | 0.05 | 0.1 | 0.2 | 0.3 | 0.4 | 0.6 | 0.8 | 1.0 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| Rolling counts | 31 | 15 | 5 | 3 | 2 | 2 | 4 | 9 |

### 16.4 ADC Input Multiplexer and Low Pass Filter

There are several analog input multiplexers and a low pass filter in FS98O02. We may use the analog input multiplexers and the low pass filter properly to measure the input signals well.

Table 16-5: ADC Input Multiplexer and Low Pass Filter Related Registers

| Address | Name | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Reset State | WDT Reset State |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 2Ah | NETB |  | SINL [1:0] |  | SINH [1:0] |  | SFT [2] |  | SFT [0] | ---0 00-0 | ---0 00-0 |
| 2Ch | NETD |  | SVRH [0] | SVRL | [1:0] |  | - | SLVD | SVR | u000 uu00 | u000 uu00 |



Figure 16-2: The FS98002A / B ADC Input Multiplexer and Low Pass Filter


Figure 16-3: The FS98002C ADC Input Multiplexer and Low Pass Filter

## ADC Operation

2. Get the VGG (2 times VDDP / 3V or external Power Supply).
3. Get the VDDA $(3.6 \mathrm{~V} / 2.5 \mathrm{~V})$
4. Enable the Analog Bias Circuit
5. Set $\operatorname{SINH}[2: 0]$ and SFTA[2:0] to decide the ADC positive input port signal.(Table 16-6, Table 16-7 and Table 16-8)

Table 16-6: FTIN Selection table

| SINH[1:0] | FTIN |
| :--- | :--- |
| 00 | OP1O |
| 01 | OP1P |
| 10 | AD2 |
| 11 | AD3 |

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Table 16-7: FTB selection table

| SFT[2] | FTB $^{1}$ |
| :--- | :--- |
| 0 | ADC Low Pass Filter is disabled |
| 1 | ADC Low Pass Filter is enabled |

Table 16-8: INH selection table

| SFT[0] | INH (ADC positive input port signal) |
| :--- | :--- |
| 0 | FTB |
| 1 | FTIN |

6. Set $\operatorname{SINL}[1: 0]$ to decide the ADC negative input port signal. (Table 16-9)

Table 16-9: INL selection table

| SINL[1:0] | INL for FS98002A / B | INL for FS98O02C |
| :--- | :--- | :--- |
| 00 | AD2 | AD2 |
| 01 | AD3 | AD3 |
| 10 | VRRL | AD3 |
| 11 | LVDL | LVDL |

7. Set ADG[1:0] to decide the ADC input gain. (Table 16-10)

Table 16-10: ADG selection table

| ADG[1:0] | ADC input gain |
| :--- | :--- |
| 00 | $2 / 3$ |
| 01 | 1 |
| 10 | 2 |
| 11 | $7 / 3$ |

8. Set SVRH[1:0] to decide the ADC reference voltage positive input port signal. (Table 16-11)

Table 16-11: VRH selection table

| SVRH[1:0] | VRH (ADC reference voltage positive input) |
| :--- | :--- |
| 0 | AD4 |
| 1 | VRRH |

9. Set SVRL[1:0] to decide the ADC reference voltage negative input port signal. (Table 16-12)

Table 16-12: VRL selection table

| SVRL[1:0] | VRL for FS98002A / B | VRL for FS98002C |
| :--- | :--- | :--- |
| 00 | AGND | AGND |
| 01 | AD3 | AD3 |
| 10 | VRRL | AD3 |
| 11 | VRRL | AD3 |

10. Set ADIE and GIE register flags to enable the ADC interrupt
11. Set ADEN register flag, the embedded $\Sigma-\Delta$ modulator will be enabled.
12. Set ADRST register flag, the comb filter will be enabled.
13. When the ADC interrupt happen, read the ADO[23:8] to get the ADC output.(ADO[23:22] are signed bits)

[^0]14. Set $A Z$ register flag to make the ADC positive and negative input port be internally short. Read the ADO[23:8] to get the ADC offset (The ADO should be zero if the offset is zero). Clear AZ register flag to make the ADC work normally.
16.5 OPAMP: OP1

Table 16-13: FS98002 OPAMP register table

| Address | Name | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Reset State | WDT Reset State |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| 09 h | PCK |  |  |  |  | S_CHCK[1:0] |  |  | ---0000 | ---0000 |  |
| 33 h | NETK | OP1EN |  |  |  |  |  |  |  | SOP1P[1:0] | SOP1N[0] |
| 0 | 00000000 | 00000000 |  |  |  |  |  |  |  |  |  |

## OPAMP Operation

1. Set SOP1P[1:0] to decide the OPAMP non-inverting input port signal. (Table 16-14)

Table 16-14: OP1P selection table

| SOP1P[1:0] | OP1P (OPAMP non-inverting input) |
| :--- | :--- |
| 00 | AGND |
| 01 | LVDH |
| 10 | AD2 |
| 11 | AD1 |

2. Set SOP1N[0] to decide the OPAMP inverting input port signal. (Table 16-15)

Table 16-15: OP1N selection table

| SOP1N[0] | OP1N (OPAMP inverting input) |
| :--- | :--- |
| 0 | OP1O |
| 1 | AD0 |

3. Set S_CHCK[1:0] to decide the OPAMP chopper mode.

Table 16-16: chopper mode selection table

| S_CHCK[1:0] | OPAMP chopper mode (input operation) |
| :--- | :--- |
| 00 | + Offset |
| 01 | -Offset |
| 10 | MCK/500 chopper frequency |
| 11 | MCK/1000 chopper frequency |

4. Set OP1EN to enable the OPAMP.

### 16.5.1 ADC Pre-filter

The input signal is sampled by the ADC. When the input signal has a noise with frequency higher than the sampling frequency, the noise generates low frequency noises through sampling circuit. Hence it is recommended to pass the input signal through a low pass filter, in order to get a stable ADC output.

Inside the chip, there is an internal 100k ohm resistor, which is for the construction of a low pass filter through parallel connection with an external capacitor between two pins FTB and FTC. The capacitance is normally between 10 nF and 50 nF . Please note that a larger capacitance may cause too much delay time in input signal switching. SFT [2] decides if an input signal passes the low pass filter. SFT [0] decides whether the ADC's input is the signal through a low pass filter.

### 16.5.2 ADC Input Multiplexers

The positive and negative input signal terminals of the ADC and reference voltages of the ADC can be selected by analog multiplexers and set to status specified by users. The input signal terminals of the ADC can be selected by SIN [0] and SFT [2]. The reference voltages of the ADC can be selected by SVR.

## 17. Instruction Programmer EPROM

"TBLP" is an instruction to programmer EPROM, "MOVP" is an instruction to look up table from EPROM. In this function, PT1[1] must connect to VSS, and VPP must be 12 V .

Table 17-1 FS98002 Programmer EPROM register table

| Address | Name | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Reset State | WDT Reset State |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| 2Ch | NETD | EPMAT |  |  |  | ERV | EPBLK |  |  | u000 uu00 | u000 uu00 |
| OAh | EADRH | PAR [15:8] |  |  |  | uuuu uuuu | uuuu uuuu |  |  |  |  |
| OBh | EADRL | PAR $[7: 0]$ |  |  | uuuu uuuu | uuuu uuuu |  |  |  |  |  |
| 0Ch | EDAH | EDATA $[15: 8]$ |  | uuuu uuuu | uuuu uuuu |  |  |  |  |  |  |

NETD[3] : ERV is a status flag of programming voltage detection. When VPP $=12 \mathrm{~V}, \mathrm{ERV}=1$. Otherwise ERV $=$ 0.

NETD[2] : EPBLK is a status flag of blank data checking generated by table look-up instruction MOVP. When data in EPROM address \{EADRH, EADRL\} equals to FFFFh, ELBLK=1. Otherwise ELBLK=0
NETD[7] : EPMAT is a EPROM data validation flag generated by instruction MOVP. When EPROM data in address \{EADRH, EADRL\} equals to the data in register \{EDAH, WORK\}, ELMAT=1. Otherwise ELMAT=0

EADRH[7:0] : Programmer Address MSB.
EADRL[7:0] : Programmer Address LSB.
EDAH[7:0] : Programmer Data MSB.
WORK[7:0] : Programmer Data LSB.

Note. 1. Please keep VDD $=3.2 \mathrm{~V} \pm 0.2 \mathrm{~V}$ when executing Instruction Programmer EPROM.
2. Charge Pump must be on and delayed about 100 ms for VGG=2xVDDP before the VPP connect to 12 V and executing Instruction Programmer EPROM.


Example 17-1: Programmer EPROM Example

## 18. LCD Driver

The LCD driver in FS98O02 has 4 commons and 12 segments, and the LCD can drive $4 \times 12,48$ dots LCD.

The LCD common driver waveform is show in Figure 18-1.


Figure 18-1: LCD Common Driver Waveform

Table 18-1: LCD Control Related Registers

| Address | Name | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Reset State | WDT Reset State |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 40h | LCD1 | SEG1 [3:0] |  |  |  | SEG0 [3:0] |  |  |  | uuuu uuuu | uuuu uuuu |
| 41h | LCD2 | SEG3 [3:0] |  |  |  | SEG2 [3:0] |  |  |  | uuuu uauu | uuuu uuuu |
| 42h | LCD3 | SEG5 [3:0] |  |  |  | SEG4 [3:0] |  |  |  | uuuu uuuu | uuuu uuuu |
| 43h | LCD4 | SEG7 [3:0] |  |  |  | SEG6 [3:0] |  |  |  | uuuu uuuu | uuuu uuuu |
| 44h | LCD5 | SEG9 [3:0] |  |  |  | SEG8 [3:0] |  |  |  | uuuu unuu | unuu unuu |
| 45h | LCD6 | SEG11 [3:0] |  |  |  | SEG10 [3:0] |  |  |  | unuu uuuu | unuu uuuu |
| 2Eh | NETF |  |  | EN_LCDB | LCDEN |  |  |  |  | 00000100 | 00000100 |

- LCD1~LCD6 is the LCD display data area.
- LCDEN $=1$ starts the LCD clock and then we may use the LCD. If $\operatorname{LCDEN}=0$, the $\operatorname{LCD}$ driver will be disabled.
When EN_LCDB = 1, the LCD bias circuit is enabled. When EN_LCDB $=0$, the LCD bias circuit is disabled, and the LCD driver will not function.

The LCD frame frequency is selected by internal LCDCKS [1:0] as described in Table 18-2.

Table 18-2: Setting LCD Frame Frequency

| LCDCKS [1:0] | LCD Frame Frequency |
| :--- | :--- |
| 01 | LCD Input Frequency/16 ( $\sim 31.25 \mathrm{~Hz})$ |

Here we demonstrate the way to light on the dots on the LCD. If we set LCD1 = "0110-1101b", it means we set SEG1 [3:0] = "0110b" and SEG0 [3:0] = "1101b". Then, the dots on the cross position of SEG1 and COM3, COM2 will be on; also, the dots on the cross position of SEG0 and COM4, COM3, COM1 will be on. We may
use the same method to light on the dots on the cross position of SEG0 to SEG7 and COM4 to COM1.

## 19. CPU Reset

The FS98002 CPU has three reset signals and they are external reset (RST_), low voltage reset (LVR), and watchdog time out reset. When resetting, the CPU's program counter (PC) is reset to 0 . After reset finished, the CPU starts working. Table 19-1 shows the CPU's internal registers status after reset.

Table 19-1: the CPU's Internal Registers Status after Reset

| Address | Name | Reset State | WDT Reset State |
| :---: | :---: | :---: | :---: |
| 00h | IND0 | uuuu uuuu | uuuu uuuu |
| 02h | FSR0 | uuuu uuuu | uuuu uuuu |
| 04h | STATUS | ---0 Ouuu | ---u 1uuu |
| 05h | WORK | uuuu uuuu | uuuu uuuu |
| 06h | INTF | ---0-00 | ---0 --00 |
| 07h | INTE | 0--0-00 | 0--0 --00 |
| 09h | PCK | --- 0000 | --- 0000 |
| 0Ah | EADRH | uuuu uuuu | uuuu uuuu |
| 0Bh | EADRL | uuuu uuuu | uuuu uuuu |
| 0Ch | EDAH | uuuu uuuu | uuuu uuuu |
| 0Dh | TMOUT | 00000000 | 00000000 |
| 0Eh | TMCON | 10000000 | 14000000 |
| 0Fh | TMMOD | 00000000 | 00000000 |
| 10h | ADOH | 00000000 | 00000000 |
| 11h | ADOM | 00000000 | 00000000 |
| 12h | ADOL | 00000000 | 00000000 |
| 13h | ADCON | ---- 0000 | --- 0000 |
| 1Fh | SVD | ---- ---1 | ------1 |
| 20h | PT1 | uuuu uuuu | uuuu uuuu |
| 21h | PT1EN | 0000 ---- | 0000 ---- |
| 22h | PT1PU | 00000000 | 00000000 |
| 23h | PT1MR | 00-- --00 | 00-- --00 |
| 24h | PT2 | ---- uuuu | ---- uuuu |
| 25h | PT2EN | ---- 0000 | ---- 0000 |
| 26h | PT2PU | ---- 0000 | ---- 0000 |
| 2Ah | NETB | -000 00-0 | -000 00-0 |
| 2Ch | NETD | u000 uu00 | u000 uu00 |
| 2Eh | NETF | 00000100 | 00000100 |
| 2Fh | NETG | ---- 0000 | ---- 0000 |
| 33h | NETK | 00000000 | 00000000 |
| 40h | LCD1 | unuu uuuu | uuuu uuuu |
| 41h | LCD2 | uuuu uuuu | uuuu uuuu |
| 42h | LCD3 | uuuu uuuu | uuuu uuuu |
| 43h | LCD4 | uuuu uuuu | uuuu uuuu |
| 44h | LCD5 | uuuu uuuu | uuuu uuuu |
| 45h | LCD6 | uuuu uuuu | uuuu uuuu |

Note 1: "u" means unknown or unchanged. "-" means unimplemented, read as " 0 ".
Note 2: The "Reset State" indicates the registers state after external reset and low voltage reset.
Note 3: The "WDT Reset State" indicates the registers state after watchdog time out reset.

### 19.1 External Reset

The CPU has a "VPP/RST/TST" pin for external reset usage. When "VPP/RST/TST" is in logic "low" state
(about $0 \sim 0.3 \mathrm{~V}$ ), the CPU will go into external reset status. The external $\mathrm{R} / \mathrm{C}$ circuit for reset is shown as following. When VDD changes from "low" to "high", the CPU external reset status will be released, and the CPU will be in normal operating condition.

The signal from the "VPP/RST/TST" pin to CPU should remain in logic "low" state for more than $2 \mu \mathrm{~s}$ to reset the CPU. If the signal from the "VPP/RST/TST" pin to CPU is in "low" state less than $2 \mu \mathrm{~s}$, the CPU will not be reset. Figure 19-2 shows the minimum reset period to reset the CPU.


Figure 19-1: the Reset Circuit and the Reset Timing

### 19.2 Low Voltage Reset

To avoiding the CPU in an abnormal power status that makes the CPU unable to reset and causes the CPU operating abnormally, there's a low voltage reset circuit embedded in FS98O02. When the voltage of VDD is less than LVR threshold low voltage, the CPU enters reset state; and when the voltage of VDD comes back above the LVR threshold high voltage, the CPU will be in normal operating condition.

### 19.3 Watchdog Time Out Reset

The watchdog timer in FS98O02 is usually used to monitor if the CPU is in normal operation. If the CPU is not in normal operation, we may use the watchdog timer to raise a watchdog time our reset and make the CPU to be back in normal operation. The watchdog timer may be used to some period wakeup-and-measuring applications to save the power for some long time monitoring portable applications.

Table 19-2: Watchdog Time Out Reset Related Registers

| Address | Name | Bit 7 | Bit 6 | Bit 5 | Bit 4 | Bit 3 | Bit 2 | Bit 1 | Bit 0 | Reset State | WDT Reset State |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| 04h | STATUS |  |  |  |  | TO |  |  |  | ---0 0uuu | ---u 1uuu |
| 0Eh | TMCON |  | WDTEN | WTS [1:0] |  |  |  |  | 10000000 | $1 u 000000$ |  |



Figure 19-3: the Block Diagram of Watchdog Timer

When starting watch dog timer (WDT), it is necessary to set WDTEN bit. Once WDTEN be set, that can't be clear by any program instruction except hardware reset (RST PIN Set to low). When the CPU executes CLRWDT instruction, the WDT counter may be reset. The clock input of the WDT comes from an internal independent R/C oscillator. The WDT output can be selected by WTS, as shown in the following table. When the WDT outputs the logic "high", the CPU will enter reset status and set TO bit to 1.

Table 19-3: Setting the Frequency of WDTOUT

| Typical Frequency of WDTOSC is about 1kHz. |  |  |  |
| :--- | :--- | :--- | :---: |
| WTS [1:0] | Frequency of WDTOUT | Typical Frequency of WDTOUT |  |
| 00 | FWDTOSC / 4096 | 0.244 Hz |  |
| 01 | FWDTOSC / 2048 | 0.488 Hz |  |
| 10 | FWDTOSC / 1024 | 0.977 Hz |  |
| 11 | FWDTOSC / 512 | 1.953 Hz |  |

## 20. Halt and Sleep Mode

### 20.1 Halt Mode

After the CPU executes a HALT instruction, the CPU program counter (PC) stops counting until the CPU receives an internal or external interrupt signal. To avoid program errors caused by Interrupt return, it is necessary to add a NOP instruction after the HALT instruction to guarantee the program's normal execution as described in Example20-1.

## HALT <br> NOP

## Example20-1: Halt Mode

### 20.2 Sleep Mode

After the CPU executes a SLEEP instruction, all oscillators stop working until the CPU receives an external interrupt signal or the CPU is reset. To avoid program errors caused by Interrupt return, it is necessary to add a NOP instruction after the SLEEP instruction to guarantee the program's normal execution as described in Example 20-2.

## SLEEP <br> NOP

## Example 20-2: Sleep Mode

To make sure that the CPU have the minimum power consumption in SLEEP mode, it is necessary to disable all the power management and analog circuits before executing a SLEEP instruction, and to make sure all the I/O ports are in VDD or VSS voltage levels. There are some parasitic diodes between VDDA and analog input ports as in Figure 20-1. When the voltage regulator is disabled and VDD is gradually going to VSS voltage
level, it is necessary to keep AD0~AD4 in floating or VSS voltage level.


Figure 20-1: the Parasitic Diodes between VDD and Analog Input Ports

Example 20-3 is a recommended example program for user's reference before the CPU executes the SLEEP instruction.


Example 20-3: Example Program before the CPU Executes the SLEEP Instruction

## 21. Instruction Set

The FS98O02 instruction set consists of 37 instructions. Each instruction is a 16-bit word with an OPCODE and one or more operands. The detail descriptions are as below.
21.1 Instruction Set Summary

Table 21-1: Instruction Set Summary Table

| Instruction | Operation | Cycle | Flag |
| :--- | :--- | :--- | :--- |
| ADDLW k | $[\mathrm{W}] \leftarrow[\mathrm{W}]+\mathrm{k}$ | 1 | $\mathrm{C}, \mathrm{DC}, \mathrm{Z}$ |
| ADDPCW | $[\mathrm{PC}] \leftarrow[\mathrm{PC}]+1+[\mathrm{W}]$ | 2 | None |
| ADDWF f, d | $[$ Destination $] \leftarrow[\mathrm{f}]+[\mathrm{W}]$ | 1 | $\mathrm{C}, \mathrm{DC}, \mathrm{Z}$ |
| ADDWFC f, d | $[$ Destination $] \leftarrow[\mathrm{f}]+[\mathrm{W}]+\mathrm{C}$ | 1 | $\mathrm{C}, \mathrm{DC}, \mathrm{Z}$ |
| ANDLW k | $[\mathrm{W}] \leftarrow[\mathrm{W}]$ AND k | 1 | Z |
| ANDWF f, d | $[$ Destination $] \leftarrow[\mathrm{W}]$ AND $[\mathrm{f}]$ | 1 | Z |
| BCF f, b | $[\mathrm{f}<\mathrm{b}>] \leftarrow 0$ | 1 | None |

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| Instruction | Operation | Cycle | Flag |
| :---: | :---: | :---: | :---: |
| BSF f, b | [ $\mathrm{f}<\mathrm{b}>$ ] $\leftarrow 1$ | 1 | None |
| BTFSC f, b | Skip if [ $f<b$ >] = 0 | 1,2 | None |
| BTFSS f, b | Skip if [ $f<b>$ ] = 1 | 1,2 | None |
| CALL k | Push PC + 1 and GOTO k | 2 | None |
| CLRF f | [f] $\leftarrow 0$ | 1 | Z |
| CLRWDT | Clear watch dog timer | 1 | None |
| COMF f, d | [f] $\leftarrow \mathrm{NOT}$ ([f]) | 1 | Z |
| DECF f, d | [Destination] $\leftarrow[f]-1$ | 1 | Z |
| DECFSZ f, d | [Destination] $\leftarrow[\mathrm{f}]-1$, skip if the result is zero | 1,2 | None |
| GOTO k | $\mathrm{PC} \leftarrow \mathrm{k}$ | 2 | None |
| HALT | CPU Stop | 1 | None |
| INCF f, d | [Destination $] \leftarrow[\mathrm{f}]+1$ | 1 | Z |
| INCFSZ f, d | [Destination] $\leftarrow[f]+1$, skip if the result is zero | 1,2 | None |
| IORLW k | $[\mathrm{W}] \leftarrow[\mathrm{W}] \mid \mathrm{k}$ | 1 | Z |
| IORWF f, d | $[$ Destination $] \leftarrow[\mathrm{W}] \mid[f]$ | 1 | Z |
| MOVFW f | $[\mathrm{W}] \leftarrow[\mathrm{f}]$ | 1 | None |
| MOVLW k | [W] $\leftarrow \mathrm{k}$ | 1 | None |
| MOVWF f | [f] $\leftarrow$ [W] | 1 | None |
| NOP | No operation | 1 | None |
| RETFIE | Pop PC and GIE = 1 | 2 | None |
| RETLW k | RETURN and W = k | 2 | None |
| RETURN | Pop PC | 2 | None |
| RLF f, d | [Destination<n+1>] $\leftarrow[f<n>]$ | 1 | C,Z |
| RRF f, d | [Destination<n-1>] $\leftarrow[\mathrm{f}<\mathrm{n}>$ ] | 1 | C, Z |
| SLEEP | Stop OSC | 1 | PD |
| SUBLW k | [W] $\leftarrow \mathrm{k}-\mathrm{LW}]$ | 1 | C, DC, Z |
| SUBWF f, d | [Destination] $\leftarrow[f]-[W]$ | 1 | C, DC, Z |
| SUBWFC f, d | $[$ Destination $] \leftarrow[f]-[\mathrm{W}]-\dot{\mathrm{C}}$ | 1 | C, DC, Z |
| XORLW k | [W] $\leftarrow[\mathrm{W}]$ XOR k | 1 | Z |
| $\begin{aligned} & \hline \text { XORWF f, d } \\ & \text { TBLP } \quad \text { k } \\ & \text { MOVP } \end{aligned}$ | ```[Destination]}\leftarrow[W] XOR [f]  [EADRH,EADRL] < EDAH,WORK (k is const) [EADRH,EADRL] }->\mathrm{ EDAH,WORK ; (EADRH,EADRL) + k``` | $\begin{array}{\|l} \hline 1 \\ \left(k^{*} 2\right)+1 \\ 2 \\ \hline \end{array}$ | Z <br> None <br> None |

- Note
- f: memory address. f may be 00h to 7Fh.
- W: work register.
- k: literal field, constant data or label.
- $d$ : destination select. If $d=0$, store result in $W$. If $d=1$, store result in memory address $f$.
- b: bit select. b may be 0 to 7 .
- [f]: the content of memory address $f$.
- PC: program counter.
- C: Carry flag.
- DC: Digit carry flag
- Z: Zero flag.
- PD: power down flag.
- TO: watchdog time out flag
- WDT: watchdog timer counter.


### 21.2 Instruction Description

The instruction descriptions are sort by alphabetically.

| ADDLW | Add Literal to W |
| :--- | :--- |
| Syntax | ADDLW k <br> $0 \leq \mathrm{k} \leq \mathrm{FFh}$ |
| Operation | $[\mathrm{W}] \leftarrow[\mathrm{W}]+\mathrm{k}$ |
| Flag Affected | C, DC, Z |
| Description | The content of Work register <br> add literal " k " in Work <br> register |
| Cycle | 1 |
| Example: | Before instruction: <br> $\mathrm{W}=08 \mathrm{~h}$ |
| ADDLW 08h | After instruction: <br> $\mathrm{W}=10 \mathrm{~h}$ |


| ADDPCW | Add W to PC |
| :--- | :--- |
| Syntax | ADDPCW |
| Operation | $[\mathrm{PC}] \leftarrow[\mathrm{PC}]+1+[\mathrm{W}],[\mathrm{W}]<$ |
|  | 79 h |
|  | $[\mathrm{PC}] \leftarrow[\mathrm{PC}]+1+([\mathrm{W}]-$ |
|  | $100 \mathrm{~h})$, otherwise |
| Flag Affected | None |
| Description | The relative address PC + 1 |
|  | +W are loaded into PC. |
| Cycle | 2 |
| Example 1: | Before instruction: |
| ADDPCW | $\mathrm{W}=7 \mathrm{Fh}, \mathrm{PC}=0212 \mathrm{~h}$ |
|  | After instruction: |
|  | $\mathrm{PC}=0292 \mathrm{~h}$ |
| Example 2: | Before instruction: |
| ADDPCW | $\mathrm{W}=80 \mathrm{~h}, \mathrm{PC}=0212 \mathrm{~h}$ |
|  | After instruction: |
|  | $\mathrm{PC}=0193 \mathrm{~h}$ |
| Example 3: | Before instruction: |
| ADDPCW | $\mathrm{W}=\mathrm{FEh}, \mathrm{PC}=0212 \mathrm{~h}$ |
|  | After instruction: |
|  | $\mathrm{PC}=0211 \mathrm{~h}$ |
|  |  |


| ADDWF | Add W to $f$ |
| :--- | :--- |
| Syntax | ADDWF f, <br> $0 \leq \leq \mathrm{f} \leq \mathrm{FFh}$ <br> $\mathrm{d} \in[0,1]$ |
| Operation | $[$ Destination $] \leftarrow[f]+[\mathrm{W}]$ |
| Flag Affected | $\mathrm{C}, \mathrm{CD}, \mathrm{Z}$ |
| Description | Add the content of the W <br> register and [f]. If $d$ is 0, the <br> result is stored in the W <br> register. If $d$ is 1, the result is <br> stored back in f. |
| Cycle | 1 |
| Example 1: | Before instruction: <br> OPERAND $=\mathrm{C} 2 \mathrm{~h}$ |



| ADDWFC | Add W, f and Carry |
| :---: | :---: |
| Syntax | $\begin{aligned} & \text { ADDWFC f, d } \\ & 0 \leq f \leq F F h \\ & d \in[0,1] \end{aligned}$ |
| Operation | [Destination $] \leftarrow[\mathrm{f}]+[\mathrm{W}]+\mathrm{C}$ |
| Flag Affected | C, DC, Z |
| Description | Add the content of the W register, [f] and Carry bit. If $d$ is 0 , the result is stored in the W register. <br> If $d$ is 1 , the result is stored back in $f$. |
| Cycle | $1 \longrightarrow$ |
| Example | Before instruction: |
| ADDWFC | C = 1 |
| OPERAND,1 | $\begin{aligned} & \text { OPERAND }=02 \mathrm{~h} \\ & W=4 D h \end{aligned}$ |

After instruction:
$C=0$
OPERAND $=50 \mathrm{~h}$
$W=4 \mathrm{Dh}$

| ANDLW | AND literal with W |
| :---: | :---: |
| Syntax | ANDLW k |
|  | $0 \leq \mathrm{k} \leq \mathrm{FFh}$ |
| Operation | [W] $\leftarrow$ [W] AND k |
| Flag Affected | Z |
| Description | AND the content of the W register with the eight-bit literal " $k$ ". <br> The result is stored in the W register. |
| Cycle | 1 |
| Example: <br> ANDLW 5Fh | Before instruction: W = A3h <br> After instruction: W = 03h |


| ANDWF | AND W and f |
| :--- | :--- |
| Syntax | ANDWF $\mathrm{f}, \mathrm{d}$ |
| $0 \leq \mathrm{f} \leq \mathrm{FFh}$ |  |
| $\mathrm{d} \in[0,1]$ |  |


| Flag Affected | $Z$ |
| :--- | :--- |
| Description | AND the content of the W |
| register with [f]. |  |
| If d is 0, the result is stored in |  |
| the W register. |  |
|  | If d is 1, the result is stored |
| back in f. |  |


| BCF | Bit Clear f |
| :---: | :---: |
| Syntax | BCF f, b |
|  | $0 \leq f \leq F F h$ |
|  | $0 \leq \mathrm{b} \leq 7$ |
| Operation | $[\mathrm{f}<\mathrm{b}>] \leftarrow 0$ |
| Flag Affected | None |
| Description | Bit b in [f] is reset to 0 . |
| Cycle | 1 |
| Example: | Before instruction: |
| BCF FLAG, 2 | FLAG = 8Dh |
|  | After instruction: FLAG = 89h |
| $\frac{\text { BSF }}{\text { Syntax }}$ | Bit Set f |
|  | BSF f, b |
|  | $0 \leq f \leq F F h$ |
|  | $0 \leq b \leq 7$ |
| Operation | $[\mathrm{f}<\mathrm{b}>$ ] $\leftarrow 1$ |
| Flag Affected | None |
| Description | Bit b in [f] is set to 1. |
| Cycle | 1 |
| Example: | Before instruction: |
| BSF FLAG, 2 | FLAG $=89 \mathrm{~h}$ |
|  | After instruction: $F L A G=8 D h$ |


| BTFSC | Bit Test skip if Clear |
| :--- | :--- |
| Syntax | BTFSC f, <br> $0 \leq f \leq F F h$ <br> $0 \leq b \leq 7$ |
| Operation | Skip if $[f<b>]=0$ |
| Flag Affected | None |
| Description | If bit 'b' in [f] is 0, the next <br> fetched instruction is <br> discarded and a NOP is |


|  | executed instead making it a two-cycle instruction. |
| :---: | :---: |
| Cycle | 1,2 |
| Example: | Before instruction: |
| Node BTFSC FLAG, | PC = address (Node) |
| 2 | After instruction: |
| OP1 | If $\mathrm{FLAG}<2>=0$ |
| OP2 | $\mathrm{PC}=$ address(OP2) |
|  | If $\mathrm{FLAG}<2>=1$ |
|  | $\mathrm{PC}=\operatorname{address}(\mathrm{OP} 1)$ |


| BTFSS | Bit Test skip if Set |
| :--- | :--- |
| Syntax | BTFSS f, $b$ |
|  | $0 \leq f \leq F F h$ |
| $0 \leq b \leq 7$ |  |
| Operation | Skip if $[f<b>]=1$ |
| Flag Affected | None |
| Description | If bit ' $b$ ' in [f] is 1, the next |

fetched instruction is
discarded and a NOP is
executed instead making it a two-cycle instruction.

|  | two-cycle instruction. |
| :--- | :--- |
| Cycle | 1,2 |
| Example: | Before instruction: |
| Node BTFSS FLAG, | PC $=$ address (Node) |
| $\mathbf{2}$ | After instruction: |
| OP1 $:$ | If FLAG $<2>=0$ |
| OP2 $\quad$ PC $=$ address(OP1) |  |
|  | If FLAG $<2>=1$ |
|  | PC $=$ address(OP2) |



| CLRF | Clear f |
| :--- | :--- |
| Syntax | CLRF f <br> $0 \leq \mathrm{f} \leq 255$ |
| Operation | $[\mathrm{f}] \leftarrow 0$ |
| Flag Affected | None |

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| Description | Reset the content of memory <br> address f |
| :--- | :--- |
| Cycle | 1 |
| Example: | Before instruction: <br> WORK = 5Ah |
| CLRF WORK | After instruction: <br> WORK $=00 \mathrm{~h}$ |


| CLRWDT | Clear watch dog timer |
| :--- | :--- |
| Syntax | CLRWDT |
| Operation | Watch dog timer counter will <br> be reset |
| Flag Affected | None |
| Description | CLRWDT instruction will <br> reset watch dog timer <br> counter. |
| Cycle | 1 |
| Example: <br> CLRWDT | After instruction: <br> WDT $=0$ |


| DECFSZ | Decrement f, skip if zero |
| :---: | :---: |
| Syntax | $\begin{aligned} & \text { DECFSZ f, } d \\ & 0 \leq f \leq F F h \\ & d \in[0,1] \end{aligned}$ |
| Operation | [Destination] $\leftarrow[f]-1$, skip if the result is zero |
| Flag Affected | None |
| Description | [ $f$ ] is decremented. If d is 0 , the result is stored in the W register. If d is 1 , the result is stored back in [f]. <br> If the result is 0 , then the next fetched instruction is discarded and a NOP is executed instead making it a two-cycle instruction. |
| Cycle | 1,2 |
| Example: Node <br> DECFSZ <br> FLAG, 1 <br> OP1 <br> OP2 | Before instruction: |
|  | PC = address (Node) |
|  | After instruction: |
|  | [FLAG] = [FLAG] - 1 |
|  | If [FLAG] $=0$ |
|  | PC = address(OP1) |
|  | If [FLAG] $\neq 0$ |
|  | PC = address(OP2) |

## Example 2:

COMF OPERAND, $1 \quad \begin{aligned} & \mathrm{W}=88 \mathrm{~h}, \mathrm{OPERAND}= \\ & \mathrm{W}\end{aligned}$ 23h


| GOTO | Unconditional Branch |
| :--- | :--- |
| Syntax | GOTO k <br> $0 \leq \mathrm{k} \leq 1 \mathrm{FFFh}$ |
| Operation | $\mathrm{PC} \leftarrow \mathrm{k}$ |
| Flag Affected | None |
| Description | The immediate address is <br> loaded into PC. |
| Cycle | 2 |
| Example: | After instruction: |
| GOTO THERE | $\mathrm{PC}=$ address (THERE) |


| HALT | Stop CPU Core Clock |
| :--- | :--- |
| Syntax | HALT |
| Operation | CPU Stop |


| Flag Affected | None | IORLW | Inclusive OR literal with W |
| :---: | :---: | :---: | :---: |
| Description | CPU clock is stopped. Oscillator is running. CPU can be waked up by internal and external interrupt sources. | Syntax | IORLW k |
|  |  |  | $0 \leq \mathrm{k} \leq \mathrm{FFh}$ |
|  |  | Operation | $[\mathrm{W}] \leftarrow[\mathrm{W}] \mid k$ |
|  |  | Flag Affected | Z |
|  |  | Description | Inclusive OR the content of the W register and the eight-bit literal "k". The result is stored in the W register. |
| Cycle | 1 |  |  |
|  |  | Cycle | 1 |
| INCF | Increment f | Example:IORLW$85 \mathbf{H}$ | Before instruction: $W=69 h$ <br> After instruction: W = EDh |
| Syntax | INCF f, d |  |  |
|  | $0 \leq f \leq F F h$ |  |  |
|  | $d \in[0,1]$ |  |  |
| Operation | [Destination] $\leftarrow[\mathrm{f}]+1$ |  |  |
| Flag Affected | Z |  |  |
| Description | [ $f$ ] is incremented. If $d$ is 0 , the result is stored in the W register. If $d$ is 1 , the result is stored back in [f]. |  |  |
|  |  | IORWF | Inclusive OR W with f |
|  |  | Syntax |  |
| Cycle | 1 |  | $\begin{aligned} & 0 \leq f \leq F F h \\ & d \in[01] \end{aligned}$ |
| Example 1: <br> INCF OPERAND, 0 | Before instruction: $\mathrm{W}=88 \mathrm{~h}$, OPERAND $=$ 23h <br> After instruction: $\mathrm{W}=24 \mathrm{~h}, \mathrm{OPERAND}=23 \mathrm{~h}$ | Operation <br> Flag Affected <br> Description | $[$ Destination $] \leftarrow[\mathrm{W}] \mid[f]$ |
|  |  |  | Z |
|  |  |  | Inclusive OR the content of the $W$ register and [ $f$ ]. If $d$ is |
| Example 2: <br> INCF OPERAND,1 | $\begin{aligned} & \text { Before instruction: } \\ & W=88 \mathrm{~h}, \text { OPERAND }= \\ & 23 \mathrm{~h} \end{aligned}$ |  | 0 , the result is stored in the $W$ register. If $d$ is 1 , the result is stored back in [ f ]. |
|  |  | Cycle Example: IORWF OPERAND,1 |  |
|  | $\begin{aligned} & W=88 \mathrm{~h}, \text { OPERAND }= \\ & 24 \mathrm{~h} \end{aligned}$ |  | Before instruction: $\mathrm{W}=88 \mathrm{~h}$, OPERAND $=$ 23h |
| INCFSZ | Increment f, skip if zero |  | After instruction: $\mathrm{W}=88 \mathrm{~h}$, OPERAND $=$ ABh |
| Syntax | INCFSZ f, d |  |  |
|  | $\begin{aligned} & 0 \leq f \leq \text { FFh } \\ & d \in[0,1] \end{aligned}$ |  |  |
|  | $\mathrm{d} \in[0,1]$ | MOVFW | Move f to W |
| Operation | [Destination] $\leftarrow[f]+1$, skip if the result is zero | Syntax | MOVFW f |
| Flag Affected | None |  | $0 \leq f \leq F F h$ |
| Description | [ $f$ ] is incremented. If $d$ is 0 , the result is stored in the W register. If $d$ is 1 , the result is stored back in [ f$]$. <br> If the result is 0 , then the next fetched instruction is discarded and a NOP is executed instead making it a two-cycle instruction. | Operation | $[\mathrm{W}] \leftarrow[\mathrm{f}]$ |
|  |  | Flag Affected | None <br> Move data from [f] to the W |
|  |  |  | register. |
|  |  | Cycle | 1 |
|  |  | Example: | Before instruction: |
|  |  | MOVFW | W = 88h, OPERAND = |
|  |  |  |  |
| Cycle | 1,2 |  | $\mathrm{W}=23 \mathrm{~h}, \mathrm{OPERAND}=$ |
| Example:NodeFLAG, 1OP1 $:$OP2 $:$ | Before instruction: <br> PC = address (Node) <br> After instruction: |  | 23h |
|  |  |  |  |  |
|  |  |  | Move literal to W |
|  | After instruction:$\begin{aligned} & {[\text { FLAG }]=[F L A G]+1} \\ & \text { If }[F L A G]=0 \\ & \text { PC }=\text { address(OP2) } \\ & \text { If }[F L A G] \neq 0 \\ & \text { PC }=\text { address(OP1) } \end{aligned}$ |  |  |
|  |  | Syntax | $0 \leq \mathrm{k} \leq \mathrm{FFh}$ |
|  |  | Operation | [W] $\leftarrow \mathrm{k}$ |
|  |  | Flag Affected | None |
|  |  | Description | Move the eight-bit literal "k" |


|  | to the content of the W <br> register. |
| :--- | :--- |
| Cycle | 1 |
| Example: | Before instruction: <br> W = 88h <br> MOVLW $\quad 23 \mathrm{H}$ <br> After instruction: <br> W = 23h |


| MOVWF | Move W to f |
| :---: | :---: |
| Syntax | MOVWF f |
|  | $0 \leq f \leq F F h$ |
| Operation | $[f] \leftarrow[\mathrm{W}]$ |
| Flag Affected | None |
| Description | Move data from the W register to [f]. |
| Cycle | 1 |
| Example: | Before instruction: |
| MOVWF OPERAND | $\begin{aligned} & \mathrm{W}=88 \mathrm{~h}, \mathrm{OPERAND}= \\ & 23 \mathrm{~h} \end{aligned}$ |
|  | After instruction: $W=88 \mathrm{~h}, \text { OPERAND }=$ 88h |


| NOP | No Operation |
| :--- | :--- |
| Syntax | NOP |
| Operation | No Operation |
| Flag Affected | None |
| Description | No operation. NOP is used <br> for one instruction cycle <br> delay. |
| Cycle | 1 |
| RETFIE | Return from Interrupt |
| Syntax | RETFIE |
| Operation | Pop Stack] => PC <br> Pop Stack <br> $1=>~ G I E ~$ |
| None |  |
| Flag Affected | The program counter is <br> loaded from the top stack, <br> then pop stack. Setting the <br> GIE bit enables interrupts. |
| Description | 2 |



| RETLW | Return and move literal to W |
| :--- | :--- |
| Syntax | RETLW k |
| $0 \leq \mathrm{k} \leq \mathrm{FFh}$ |  |
| Operation | $[$ W] $\leftarrow \mathrm{k}$ |
|  | $[$ Top Stack $]=>$ PC |


|  | Pop Stack |
| :--- | :--- |
| Flag Affected | None |
| Description | Move the eight-bit literal "k" <br> to the content of the W <br> register. The program <br> counter is loaded from the <br> top stack, then pop stack. |
| Cycle | 2 |
| Example: | Before instruction: <br> WREG $=0 \times 07$ |
| CALL TABLE | After instruction: <br> WREG $=$ value of k7 |


| Return | Return from Subroutine |
| :--- | :--- |
| Syntax | RETURN |
| Operation | $[$ Top Stack] => PC <br> Pop Stack |
| Flag Affected | None |
| Description | The program counter is <br> loaded from the top stack, <br> then pop stack. |
| Cycle | 2 |
| Example: | After instruction: <br> Return |

After instruction:
$C=1$
$\mathrm{W}=88 \mathrm{~h}$,
OPERAND $=\mathrm{CCh}$

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Package Information


NDTES:
1, WEDES DUTLINE:MS-D26 BCD
2.DIWENSIDNS D1 AND E1 DO NOT INELUDE
 0.25 mm PER SIDE. D1 AND E1 ARE HAXIWLW $0.25 m m$ PER SIDE. D1 AND E1 ARE MAXIWH
PLAלTIC 日COT SIZE DIMENSIDNS INELUDING PLASTE BDTT 51
WOLD WISWARTCH.


LQFP 64
 PROTRUSIONALLOWMEBE DAWBFR PROTRUSION SHALL NOT CAUSE THE EADD MDTH TO EXCEES THE MAXIMUM b DIMENSION BY MORE THAN DOBmm.

## Figure 22-1: LQFP64 Package Outline

## 22. Ordering Information

| Product Number | Package Type |
| :--- | :--- |
| FS98O02 | Die form (49-pin), 64-pin LQFP (Green package) |

## 23. Revision History

| Version | Date | Page | Description |
| :---: | :---: | :---: | :--- |
| 1.0 | $2010 / 06 / 21$ | All | Officially released version 1.0. |
| 1.1 | $2010 / 8 / 25$ | 11 | The op offset value changed from 1mv to 1.5 mv. |
| 1.2 | $2010 / 10 / 29$ | 10 | 4MHz FRC Minimum value modify from 3.2MHz to 3.0MHz. <br> 2.5 V VDDA Minimum value modify from 2.4V to 2.37V. |
| 1.3 | $2011 / 09 / 29$ | All | Add the FS98O02C Function Spec |
| 1.4 | $2014 / 01 / 14$ | 14 | Revise OPAMP Characteristics |
| 1.5 | $2014 / 05 / 22$ | 2 | Revised company address |


[^0]:    ${ }^{1}$ The input of ADC Low Pass Filter is FTIN, and the output is FTB

